#### **Features**

- High-performance, Low-power 8/16-bit AVR XMEGA Microcontroller
- Non-volatile Program and Data Memories
  - 16 KB 128 KB of In-System Self-Programmable Flash
  - 4 KB 8 KB Boot Code Section with Independent Lock Bits
  - 1 KB 2 KB EEPROM
  - 2 KB 8 KB Internal SRAM
- Peripheral Features
  - Four-channel DMA Controller with support for external requests
  - Eight-channel Event System
  - Five 16-bit Timer/Counters

Three Timer/Counters with 4 Output Compare or Input Capture channels Two Timer/Counters with 2 Output Compare or Input Capture channels High-Resolution Extensions on all Timer/Counters

Advanced Waveform Extension on one Timer/Counter

- Five USARTs

IrDA Extension on one USART

- Two Two-Wire Interfaces with dual address match (I<sup>2</sup>C and SMBus compatible)
- Two SPIs (Serial Peripheral Interfaces) peripherals
- AES and DES Crypto Engine
- 16-bit Real Time Counter with Separate Oscillator
- One Twelve-channel, 12-bit, 2 Msps Analog to Digital Converter
- One Two-channel, 12-bit, 1 Msps Digital to Analog Converter
- Two Analog Comparators with Window compare function
- External Interrupts on all General Purpose I/O pins
- Programmable Watchdog Timer with Separate On-chip Ultra Low Power Oscillator
- Special Microcontroller Features
  - Power-on Reset and Programmable Brown-out Detection
  - Internal and External Clock Options with PLL
  - Programmable Multi-level Interrupt Controller
  - Sleep Modes: Idle, Power-down, Standby, Power-save, Extended Standby
  - Advanced Programming, Test and Debugging Interfaces

PDI (Program and Debug Interface) for programming, test and debugging

- I/O and Packages
  - 34 Programmable I/O Lines
  - 44 lead TQFP
  - 44 pad VQFN/QFN
  - 49 ball VFBGA
- **Operating Voltage** 
  - 1.6 3.6V
- Speed performance
  - 0 12 MHz @ 1.6 3.6V
  - 0 32 MHz @ 2.7 3.6V

# Typical Applications

- Industrial control
- Climate control

- Factory automation
- ZigBee

- Building control
- Motor control Networking

- Board control White Goods
- Optical
- Hand-held battery applications
- Power tools
- HVAC
- Metering
- Medical Applications



# 8/16-bit **AVR** XMEGA A4 Microcontroller

ATxmega128A4 ATxmega64A4 ATxmega32A4 ATxmega16A4

**Preliminary** 



8069L-AVR-11/09

# **Ordering Information**

Ordering Code	Flash	E <sup>2</sup>	SRAM	Speed (MHz)	Power Supply	Package <sup>(1)(2)(3)</sup>	Temp
ATxmega128A4-AU	128 KB + 8 KB	2 KB	8 KB	32	1.6 - 3.6V		
ATxmega64A4-AU	64 KB + 4 KB	2 KB	4 KB	32	1.6 - 3.6V	44A	
ATxmega32A4-AU	32 KB + 4 KB	1 KB	4 KB	32	1.6 - 3.6V	44A	
ATxmega16A4-AU	16 KB + 4 KB	1 KB	2 KB	32	1.6 - 3.6V		
ATxmega128A4-MH	128 KB + 8 KB	2 KB	8 KB	32	1.6 - 3.6V		-40°C - 85°C
ATxmega64A4-MH	64 KB + 4 KB	2 KB	4 KB	32	1.6 - 3.6V	44M1	-40 C - 65 C
ATxmega32A4-MH	32 KB + 4 KB	1 KB	4 KB	32	1.6 - 3.6V	441011	
ATxmega16A4-MH	16 KB + 4 KB	1 KB	2 KB	32	1.6 - 3.6V		
ATxmega32A4-CU	32 KB + 4K	1 KB	4 KB	32	1.6 - 3.6V	49C2	
ATxmega16A4-CU	16 KB + 4 KB	1 KB	2 KB	32	1.6 - 3.6V	4902	

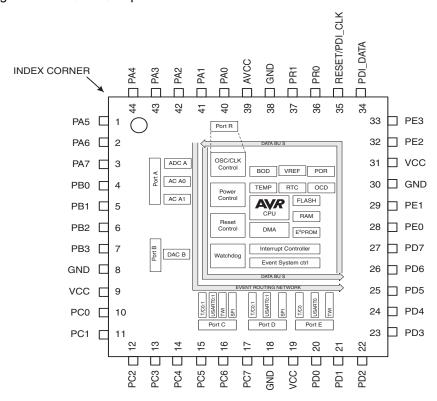
- Notes: 1. This device can also be supplied in wafer form. Please contact your local Atmel sales office for detailed ordering information.
  - 2. Pb-free packaging, complies to the European Directive for Restriction of Hazardous Substances (RoHS directive). Also Halide free and fully Green.
  - 3. For packaging information see "Packaging information" on page 58.

	Package Type						
44A	44-Lead, 10 x 10 mm Body Size, 1.0 mm Body Thickness, 0.8 mm Lead Pitch, Thin Profile Plastic Quad Flat Package (TQFP)						
44M1	44-Pad, 7x7x1 mm Body, Lead Pitch 0.50 mm, 5.20 mm Exposed Pad, Thermally Enhanced Plastic Very Thin Quad No Lead Package (VQFN)						
49C2	49-Ball (7 x 7 Array), 0.65 mm Pitch, 5.0 x 5.0 x 1.0 mm, Very Thin, Fine-Pitch Ball Grid Array Package (VFBGA)						



# 2. Pinout/Block Diagram

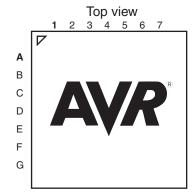
Figure 2-1. Bock Diagram and TQFP/QFN pinout



Note: For full details on pinout and pin functions refer to "Pinout and Pin Functions" on page 49.



Figure 2-2. VFBGA pinout



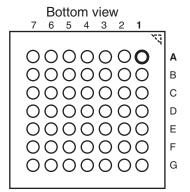


Table 2-1.VFBGA pinout

	1	2	3	4	5	6	
Α	PA3	AVCC	GND	PR1	PR0	PDI_DATA	PE3
В	PA4	PA1	PA0	GND	RESET/ PDI_CLK	PE2	VCC
С	PA5	PA2	PA6	PA7	GND	PE1	GND
D	PB1	PB2	PB3	PB0	GND	PD7	PE0
E	GND	GND	PC3	GND	PD4	PD5	PD6
F	VCC	PC0	PC4	PC6	PD0	PD1	PD3
G	PC1	PC2	PC5	PC7	GND	VCC	PD2



### 3. Overview

The XMEGA<sup>™</sup>A4 is a family of low power, high performance and peripheral rich CMOS 8/16-bit microcontrollers based on the AVR<sup>®</sup> enhanced RISC architecture. By executing powerful instructions in a single clock cycle, the XMEGA A4 achieves throughputs approaching 1 Million Instructions Per Second (MIPS) per MHz allowing the system designer to optimize power consumption versus processing speed.

The AVR CPU combines a rich instruction set with 32 general purpose working registers. All the 32 registers are directly connected to the Arithmetic Logic Unit (ALU), allowing two independent registers to be accessed in one single instruction, executed in one clock cycle. The resulting architecture is more code efficient while achieving throughputs many times faster than conventional single-accumulator or CISC based microcontrollers.

The XMEGA A4 devices provide the following features: In-System Programmable Flash with Read-While-Write capabilities, Internal EEPROM and SRAM, four-channel DMA Controller, eight-channel Event System, Programmable Multi-level Interrupt Controller, 34 general purpose I/O lines, 16-bit Real Time Counter (RTC), five flexible 16-bit Timer/Counters with compare modes and PWM, five USARTs, two Two Wire Serial Interfaces (TWIs), two Serial Peripheral Interfaces (SPIs), AES and DES crypto engine, one Twelve-channel, 12-bit ADC with optional differential input with programmable gain, one Two-channel 12-bit DAC, two analog comparators with window mode, programmable Watchdog Timer with separate Internal Oscillator, accurate internal oscillators with PLL and prescaler and programmable Brown-Out Detection.

The Program and Debug Interface (PDI), a fast 2-pin interface for programming and debugging, is available.

The XMEGA A4 devices have five software selectable power saving modes. The Idle mode stops the CPU while allowing the SRAM, DMA Controller, Event System, Interrupt Controller and all peripherals to continue functioning. The Power-down mode saves the SRAM and register contents but stops the oscillators, disabling all other functions until the next TWI or pin-change interrupt, or Reset. In Power-save mode, the asynchronous Real Time Counter continues to run, allowing the application to maintain a timer base while the rest of the device is sleeping. In Standby mode, the Crystal/Resonator Oscillator is kept running while the rest of the device is sleeping. This allows very fast start-up from external crystal combined with low power consumption. In Extended Standby mode, both the main Oscillator and the Asynchronous Timer continue to run. To further reduce power consumption, the peripheral clock to each individual peripheral can optionally be stopped in Active mode and in Idle sleep mode.

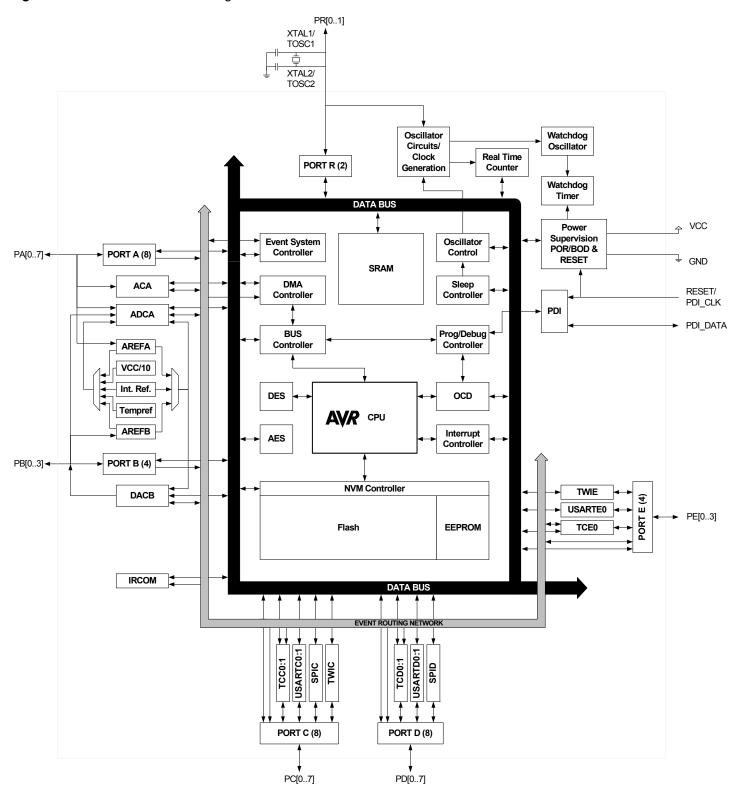
The device is manufactured using Atmel's high-density nonvolatile memory technology. The program Flash memory can be reprogrammed in-system through the PDI. A Bootloader running in the device can use any interface to download the application program to the Flash memory. The Bootloader software in the Boot Flash section will continue to run while the Application Flash section is updated, providing true Read-While-Write operation. By combining an 8/16-bit RISC CPU with In-System Self-Programmable Flash, the Atmel XMEGA A4 is a powerful microcontroller family that provides a highly flexible and cost effective solution for many embedded applications.

The XMEGA A4 devices are supported with a full suite of program and system development tools including: C compilers, macro assemblers, program debugger/simulators, programmers, and evaluation kits.



# 3.1 Block Diagram

Figure 3-1. XMEGA A4 Block Diagram





### 4. Resources

A comprehensive set of development tools, application notes and datasheets are available for download on http://www.atmel.com/avr.

### 4.1 Recommended reading

- XMEGA A Manual
- XMEGA A Application Notes

This device data sheet only contains part specific information and a short description of each peripheral and module. The XMEGA A Manual describes the modules and peripherals in depth. The XMEGA A application notes contain example code and show applied use of the modules and peripherals.

The XMEGA A Manual and Application Notes are available from http://www.atmel.com/avr.

### 5. Disclaimer

For devices that are not available yet, typical values contained in this datasheet are based on simulations and characterization of other AVR XMEGA microcontrollers manufactured on the same process technology. Min. and Max values will be available after the device is characterized.



### AVR CPU

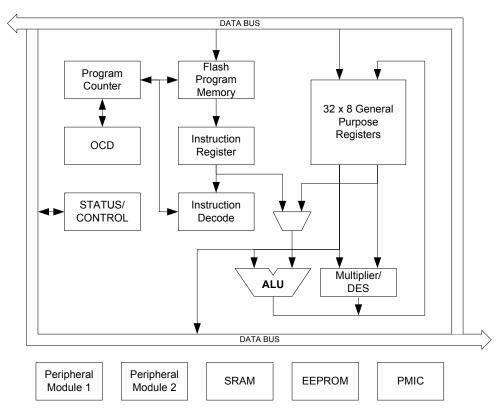
### 6.1 Features

- 8/16-bit high performance AVR RISC Architecture
  - 138 instructions
  - Hardware multiplier
- 32x8-bit registers directly connected to the ALU
- Stack in RAM
- Stack Pointer accessible in I/O memory space
- Direct addressing of up to 16M Bytes of program and data memory
- True 16/24-bit access to 16/24-bit I/O registers
- Support for 8-, 16- and 32-bit Arithmetic
- Configuration Change Protection of system critical features

#### 6.2 Overview

The XMEGA A4 uses the 8/16-bit AVR CPU. The main function of the CPU is program execution. The CPU must therefore be able to access memories, perform calculations and control peripherals. Interrupt handling is described in a separate section. Figure 6-1 on page 8 shows the CPU block diagram.

Figure 6-1. CPU block diagram



The AVR uses a Harvard architecture - with separate memories and buses for program and data. Instructions in the program memory are executed with a single level pipeline. While one instruction is being executed, the next instruction is pre-fetched from the program memory. This



concept enables instructions to be executed in every clock cycle. The program memory is In-System Re-programmable Flash memory.

### 6.3 Register File

The fast-access Register File contains 32 x 8-bit general purpose working registers with single clock cycle access time. This allows single-cycle Arithmetic Logic Unit (ALU) operation. In a typical ALU cycle, the operation is performed on two Register File operands, and the result is stored back in the Register File.

Six of the 32 registers can be used as three 16-bit address register pointers for data space addressing - enabling efficient address calculations. One of these address pointers can also be used as an address pointer for look up tables in Flash program memory.

### 6.4 ALU - Arithmetic Logic Unit

The high performance Arithmetic Logic Unit (ALU) supports arithmetic and logic operations between registers or between a constant and a register. Single register operations can also be executed. Within a single clock cycle, arithmetic operations between general purpose registers or between a register and an immediate are executed. After an arithmetic or logic operation, the Status Register is updated to reflect information about the result of the operation.

The ALU operations are divided into three main categories – arithmetic, logical, and bit-functions. Both 8- and 16-bit arithmetic is supported, and the instruction set allows for efficient implementation of 32-bit arithmetic. The ALU also provides a powerful multiplier supporting both signed and unsigned multiplication and fractional format.

### 6.5 Program Flow

When the device is powered on, the CPU starts to execute instructions from the lowest address in the Flash Program Memory '0'. The Program Counter (PC) addresses the next instruction to be fetched. After a reset, the PC is set to location '0'.

Program flow is provided by conditional and unconditional jump and call instructions, capable of addressing the whole address space directly. Most AVR instructions use a 16-bit word format, while a limited number uses a 32-bit format.

During interrupts and subroutine calls, the return address PC is stored on the Stack. The Stack is effectively allocated in the general data SRAM, and consequently the Stack size is only limited by the total SRAM size and the usage of the SRAM. After reset the Stack Pointer (SP) points to the highest address in the internal SRAM. The SP is read/write accessible in the I/O memory space, enabling easy implementation of multiple stacks or stack areas. The data SRAM can easily be accessed through the five different addressing modes supported in the AVR CPU.



### 7. Memories

#### 7.1 Features

- Flash Program Memory
  - One linear address space
  - In-System Programmable
  - Self-Programming and Bootloader support
  - Application Section for application code
  - Application Table Section for application code or data storage
  - Boot Section for application code or bootloader code
  - Separate lock bits and protection for all sections
  - Built in fast CRC check of a selectable flash program memory section
- Data Memory
  - One linear address space
  - Single cycle access from CPU
  - SRAM
  - EEPROM

Byte and page accessible

Optional memory mapping for direct load and store

- I/O Memory

Configuration and Status registers for all peripherals and modules 16 bit-accessible General Purpose Register for global variables or flags

- Bus arbitration

Safe and deterministic handling of CPU and DMA Controller priority

- Separate buses for SRAM, EEPROM, I/O Memory and External Memory access
   Simultaneous bus access for CPU and DMA Controller
- Production Signature Row Memory for factory programmed data

Device ID for each microcontroller device type

Serial number for each device

Oscillator calibration bytes

ADC, DAC and temperature sensor calibration data

User Signature Row

One flash page in size

Can be read and written from software

Content is kept after chip erase

#### 7.2 Overview

The AVR architecture has two main memory spaces, the Program Memory and the Data Memory. In addition, the XMEGA A4 features an EEPROM Memory for non-volatile data storage. All three memory spaces are linear and require no paging. The available memory size configurations are shown in "Ordering Information" on page 2. In addition each device has a Flash memory signature row for calibration data, device identification, serial number etc.

Non-volatile memory spaces can be locked for further write or read/write operations. This prevents unrestricted access to the application software.



### 7.3 In-System Programmable Flash Program Memory

The XMEGA A4 devices contain On-chip In-System Programmable Flash memory for program storage, see Figure 7-1 on page 11. Since all AVR instructions are 16- or 32-bits wide, each Flash address location is 16 bits.

The Program Flash memory space is divided into Application and Boot sections. Both sections have dedicated Lock Bits for setting restrictions on write or read/write operations. The Store Program Memory (SPM) instruction must reside in the Boot Section when used to write to the Flash memory.

A third section inside the Application section is referred to as the Application Table section which has separate Lock bits for storage of write or read/write protection. The Application Table section can be used for storing non-volatile data or application software.

Figure 7-1. Flash Program Memory (Hexadecimal address)

#### **Word Address** 0 **Application Section** (128 KB/64 KB/32 KB/16 KB) 37FF 17FF **EFFF** 77FF / F000 7800 1800 / / 3800 / **Application Table Section** (4 KB/4 KB/4 KB) **FFFF** 7FFF 3FFF 1FFF 10000 / 8000 4000 / 2000 **Boot Section** (8 KB/4 KB/4 KB/4 KB) 10FFF 87FF 47FF 27FF

The Application Table Section and Boot Section can also be used for general application software.



### 7.4 Data Memory

The Data Memory consist of the I/O Memory, EEPROM and SRAM memories, all within one linear address space, see Figure 7-2 on page 12. To simplify development, the memory map for all devices in the family is identical and with empty, reserved memory space for smaller devices.

Figure 7-2. Data Memory Map (Hexadecimal address)

Byte Address	ATxmega64A4	Byte Address	ATxmega32A4	Byte Address	ATxmega16A4
0	I/O Registers	0	I/O Registers	0	I/O Registers
FFF	(4 KB)	FFF	(4 KB)	FFF	(4 KB)
1000	EEPROM	1000	EEPROM	1000	EEPROM
17FF	(2 KB)	13FF	(1 KB)	13FF	(1 KB)
	RESERVED		RESERVED		RESERVED
2000	Internal SRAM	2000	Internal SRAM	2000	Internal SRAM
2FFF	(4 KB)	2FFF	(4 KB)	27FF	(2 KB)

Byte Address	ATxmega128A4
0	I/O Registers
FFF	(4 KB)
1000	EEPROM
17FF	(2 KB)
	RESERVED
2000	Internal SRAM
3FFF	(8 KB)

### 7.4.1 I/O Memory

All peripherals and modules are addressable through I/O memory locations in the data memory space. All I/O memory locations can be accessed by the Load (LD/LDS/LDD) and Store (ST/STS/STD) instructions, transferring data between the 32 general purpose registers in the CPU and the I/O Memory.

The IN and OUT instructions can address I/O memory locations in the range 0x00 - 0x3F directly.

I/O registers within the address range 0x00 - 0x1F are directly bit-accessible using the SBI and CBI instructions. The value of single bits can be checked by using the SBIS and SBIC instructions on these registers.

The I/O memory address for all peripherals and modules in XMEGA A4 is shown in the "Peripheral Module Address Map" on page 53.

### 7.4.2 SRAM Data Memory

The XMEGA A4 devices have internal SRAM memory for data storage.



#### 7.4.3 EEPROM Data Memory

The XMEGA A4 devices have internal EEPROM memory for non-volatile data storage. It is addressable either in a separate data space or it can be memory mapped into the normal data memory space. The EEPROM memory supports both byte and page access.

### 7.5 Production Signature Row

The Production Signature Row is a separate memory section for factory programmed data. It contains calibration data for functions such as oscillators and analog modules.

The production signature row also contains a device ID that identify each microcontroller device type, and a serial number that is unique for each manufactured device. The device ID for the available XMEGA A4 devices is shown in Table 7-1 on page 13. The serial number consist of the production LOT number, wafer number, and wafer coordinates for the device.

The production signature row can not be written or erased, but it can be read from both application software and external programming.

**Table 7-1.** Device ID bytes for XMEGA A4 devices.

Device	Device ID bytes				
	Byte 2	Byte 1	Byte 0		
ATxmega16A4	41	94	1E		
ATxmega32A4	41	95	1E		
ATxmega64A4	46	96	1E		
ATxmega128A4	46	97	1E		

### 7.6 User Signature Row

The User Signature Row is a separate memory section that is fully accessible (read and write) from application software and external programming. The user signature row is one flash page in size, and is meant for static user parameter storage, such as calibration data, custom serial numbers or identification numbers, random number seeds etc. This section is not erased by Chip Erase commands that erase the Flash, and requires a dedicated erase command. This ensures parameter storage during multiple program/erase session and on-chip debug sessions.



### 7.7 Flash and EEPROM Page Size

The Flash Program Memory and EEPROM data memory are organized in pages. The pages are word accessible for the Flash and byte accessible for the EEPROM.

Table 7-2 on page 14 shows the Flash Program Memory organization. Flash write and erase operations are performed on one page at a time, while reading the Flash is done one byte at a time. For Flash access the Z-pointer (Z[m:n]) is used for addressing. The most significant bits in the address (FPAGE) give the page number and the least significant address bits (FWORD) give the word in the page.

Devices	Flash	Page Size	FWORD	FPAGE	Appli	cation	Во	oot
	Size	(words)			Size	No of Pages	Size	No of Pages
ATxmega16A4	16 KB + 4 KB	128	Z[6:0]	Z[13:7]	16 KB	64	4 KB	16
ATxmega32A4	32 KB + 4 KB	128	Z[6:0]	Z[14:7]	32 KB	128	4 KB	16
ATxmega64A4	64 KB + 4 KB	128	Z[6:0]	Z[15:7]	64 KB	128	4 KB	16
ATxmega128A4	128 KB + 8 KB	256	<i>7</i> [7:0]	<i>7</i> [16:8]	128 KB	256	8 KB	16

**Table 7-2.** Number of words and Pages in the Flash.

Table 7-3 on page 14 shows EEPROM memory organization for the XMEGA A4 devices. EEPROM write and erase operations can be performed one page or one byte at a time, while reading the EEPROM is done one byte at a time. For EEPROM access the NVM Address Register (ADDR[m:n]) is used for addressing. The most significant bits in the address (E2PAGE) give the page number and the least significant address bits (E2BYTE) give the byte in the page.

Table 7-3.	Number of Bytes and Pages in the EEPROM.

Devices	EEPROM	Page Size	E2BYTE	E2PAGE	No of Pages
	Size	(Bytes)			
ATxmega16A4	1 KB	32	ADDR[4:0]	ADDR[10:5]	32
ATxmega32A4	1 KB	32	ADDR[4:0]	ADDR[10:5]	32
ATxmega64A4	2 KB	32	ADDR[4:0]	ADDR[10:5]	64
ATxmega128A4	2 KB	32	ADDR[4:0]	ADDR[10:5]	64



# 8. DMAC - Direct Memory Access Controller

#### 8.1 Features

- Allows High-speed data transfer
  - From memory to peripheral
  - From memory to memory
  - From peripheral to memory
  - From peripheral to peripheral
- 4 Channels
- From 1 byte and up to 16 M bytes transfers in a single transaction
- . Multiple addressing modes for source and destination address
  - Increment
  - Decrement
  - Static
- 1, 2, 4, or 8 bytes Burst Transfers
- Programmable priority between channels

### 8.2 Overview

The XMEGA A4 has a Direct Memory Access (DMA) Controller to move data between memories and peripherals in the data space. The DMA controller uses the same data bus as the CPU to transfer data.

It has 4 channels that can be configured independently. Each DMA channel can perform data transfers in blocks of configurable size from 1 to 64K bytes. A repeat counter can be used to repeat each block transfer for single transactions up to 16M bytes. Each DMA channel can be configured to access the source and destination memory address with incrementing, decrementing or static addressing. The addressing is independent for source and destination address. When the transaction is complete the original source and destination address can automatically be reloaded to be ready for the next transaction.

The DMAC can access all the peripherals through their I/O memory registers, and the DMA may be used for automatic transfer of data to/from communication modules, as well as automatic data retrieval from ADC conversions, data transfer to DAC conversions, or data transfer to or from port pins. A wide range of transfer triggers is available from the peripherals, Event System and software. Each DMA channel has different transfer triggers.

To allow for continuous transfer, two channels can be interlinked so that the second takes over the transfer when the first is finished and vice versa.

The DMA controller can read from memory mapped EEPROM, but it cannot write to the EEPROM or access the Flash.



# 9. Event System

### 9.1 Features

- · Inter-peripheral communication and signalling with minimum latency
- . CPU and DMA independent operation
- 8 Event Channels allow for up to 8 signals to be routed at the same time
- Events can be generated by
  - Timer/Counters (TCxn)
  - Real Time Counter (RTC)
  - Analog to Digital Converters (ADCx)
  - Analog Comparators (ACx)
  - Ports (PORTx)
  - System Clock (Clk<sub>SYS</sub>)
  - Software (CPU)
- · Events can be used by
  - Timer/Counters (TCxn)
  - Analog to Digital Converters (ADCx)
  - Digital to Analog Converters (DACx)
  - Ports (PORTx)
  - DMA Controller (DMAC)
  - IR Communication Module (IRCOM)
- The same event can be used by multiple peripherals for synchronized timing
- Advanced Features
  - Manual Event Generation from software (CPU)
  - Quadrature Decoding
  - Digital Filtering
- · Functions in Active and Idle mode

#### 9.2 Overview

The Event System is a set of features for inter-peripheral communication. It enables the possibility for a change of state in one peripheral to automatically trigger actions in one or more peripherals. Whose changes in a peripheral that will trigger actions in other peripherals are configurable by software. It is a simple, but powerful system as it allows for autonomous control of peripherals without any use of interrupts, CPU or DMA resources.

The indication of a change in a peripheral is referred to as an event, and is usually the same as the interrupt conditions for that peripheral. Events are passed between peripherals using a dedicated routing network called the Event Routing Network. Figure 9-1 on page 17 shows a basic block diagram of the Event System with the Event Routing Network and the peripherals to which it is connected. This highly flexible system can be used for simple routing of signals, pin functions or for sequencing of events.

The maximum latency is two CPU clock cycles from when an event is generated in one peripheral, until the actions are triggered in one or more other peripherals.

The Event System is functional in both Active and Idle modes.



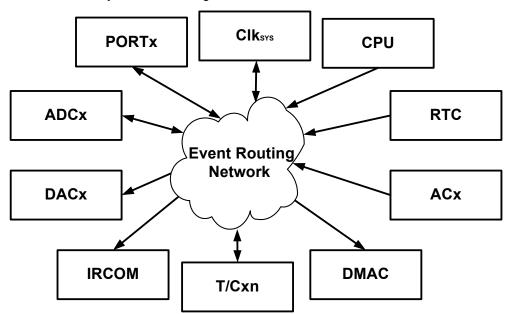


Figure 9-1. Event System Block Diagram

The Event Routing Network can directly connect together ADCs, DACs, Analog Comparators (ACx), I/O ports (PORTx), the Real-time Counter (RTC), Timer/Counters (T/C) and the IR Communication Module (IRCOM). Events can also be generated from software (CPU).

All events from all peripherals are always routed into the Event Routing Network. This consist of eight multiplexers where each can be configured in software to select which event to be routed into that event channel. All eight event channels are connected to the peripherals that can use events, and each of these peripherals can be configured to use events from one or more event channels to automatically trigger a software selectable action.



# 10. System Clock and Clock options

#### 10.1 Features

- Fast start-up time
- Safe run-time clock switching
- Internal Oscillators:
  - 32 MHz run-time calibrated RC oscillator
  - 2 MHz run-time calibrated RC oscillator
  - 32.768 kHz calibrated RC oscillator
  - 32 kHz Ultra Low Power (ULP) oscillator with 1 kHz ouput
- External clock options
  - 0.4 16 MHz Crystal Oscillator
  - 32 kHz Crystal Oscillator
  - External clock
- PLL with internal and external clock options with 1 to 31x multiplication
- Clock Prescalers with 1 to 2048x division
- Fast peripheral clock running at 2 and 4 times the CPU clock speed
- Automatic Run-Time Calibration of internal oscillators
- · Crystal Oscillator failure detection

#### 10.2 Overview

XMEGA A4 has an advanced clock system, supporting a large number of clock sources. It incorporates both integrated oscillators, external crystal oscillators and resonators. A high frequency Phase Locked Loop (PLL) and clock prescalers can be controlled from software to generate a wide range of clock frequencies from the clock source input.

It is possible to switch between clock sources from software during run-time. After reset the device will always start up running from the 2 Mhz internal oscillator.

A calibration feature is available, and can be used for automatic run-time calibration of the internal 2 MHz and 32 MHz oscillators. This reduce frequency drift over voltage and temperature.

A Crystal Oscillator Failure Monitor can be enabled to issue a Non-Maskable Interrupt and switch to internal oscillator if the external oscillator fails. Figure 10-1 on page 19 shows the principal clock system in XMEGA A4.



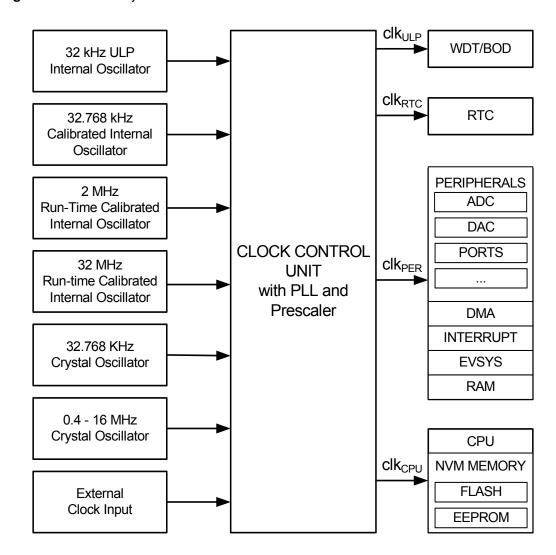


Figure 10-1. Clock system overview

Each clock source is briefly described in the following sub-sections.

### 10.3 Clock Options

### 10.3.1 32 kHz Ultra Low Power Internal Oscillator

The 32 kHz Ultra Low Power (ULP) Internal Oscillator is a very low power consumption clock source. It is used for the Watchdog Timer, Brown-Out Detection and as an asynchronous clock source for the Real Time Counter. This oscillator cannot be used as the system clock source, and it cannot be directly controlled from software.

### 10.3.2 32.768 kHz Calibrated Internal Oscillator

The 32.768 kHz Calibrated Internal Oscillator is a high accuracy clock source that can be used as the system clock source or as an asynchronous clock source for the Real Time Counter. It is calibrated during production to provide a default frequency which is close to its nominal frequency.



#### 10.3.3 32.768 kHz Crystal Oscillator

The 32.768 kHz Crystal Oscillator is a low power driver for an external watch crystal. It can be used as system clock source or as asynchronous clock source for the Real Time Counter.

### 10.3.4 0.4 - 16 MHz Crystal Oscillator

The 0.4 - 16 MHz Crystal Oscillator is a driver intended for driving both external resonators and crystals ranging from 400 kHz to 16 MHz.

#### 10.3.5 2 MHz Run-time Calibrated Internal Oscillator

The 2 MHz Run-time Calibrated Internal Oscillator is a high frequency oscillator. It is calibrated during production to provide a default frequency which is close to its nominal frequency. The oscillator can use the 32 kHz Calibrated Internal Oscillator or the 32 kHz Crystal Oscillator as a source for calibrating the frequency run-time to compensate for temperature and voltage drift hereby optimizing the accuracy of the oscillator.

#### 10.3.6 32 MHz Run-time Calibrated Internal Oscillator

The 32 MHz Run-time Calibrated Internal Oscillator is a high frequency oscillator. It is calibrated during production to provide a default frequency which is close to its nominal frequency. The oscillator can use the 32 kHz Calibrated Internal Oscillator or the 32 kHz Crystal Oscillator as a source for calibrating the frequency run-time to compensate for temperature and voltage drift hereby optimizing the accuracy of the oscillator.

### 10.3.7 External Clock input

The external clock input gives the possibility to connect a clock from an external source.

#### 10.3.8 PLL with Multiplication factor 1 - 31x

The PLL provides the possibility of multiplying a frequency by any number from 1 to 31. In combination with the prescalers, this gives a wide range of output frequencies from all clock sources.



# 11. Power Management and Sleep Modes

#### 11.1 Features

- · 5 sleep modes
  - Idle
  - Power-down
  - Power-save
  - Standby
  - Extended standby
- Power Reduction registers to disable clocks to unused peripherals

### 11.2 Overview

The XMEGA A4 provides various sleep modes tailored to reduce power consumption to a minimum. All sleep modes are available and can be entered from Active mode. In Active mode the CPU is executing application code. The application code decides when and what sleep mode to enter. Interrupts from enabled peripherals and all enabled reset sources can restore the microcontroller from sleep to Active mode.

In addition, Power Reduction registers provide a method to stop the clock to individual peripherals from software. When this is done, the current state of the peripheral is frozen and there is no power consumption from that peripheral. This reduces the power consumption in Active mode and Idle sleep mode.

### 11.3 Sleep Modes

#### 11.3.1 Idle Mode

In Idle mode the CPU and Non-Volatile Memory are stopped, but all peripherals including the Interrupt Controller, Event System and DMA Controller are kept running. Interrupt requests from all enabled interrupts will wake the device.

#### 11.3.2 Power-down Mode

In Power-down mode all system clock sources, and the asynchronous Real Time Counter (RTC) clock source, are stopped. This allows operation of asynchronous modules only. The only interrupts that can wake up the MCU are the Two Wire Interface address match interrupts, and asynchronous port interrupts, e.g pin change.

#### 11.3.3 Power-save Mode

Power-save mode is identical to Power-down, with one exception: If the RTC is enabled, it will keep running during sleep and the device can also wake up from RTC interrupts.

### 11.3.4 Standby Mode

Standby mode is identical to Power-down with the exception that all enabled system clock sources are kept running, while the CPU, Peripheral and RTC clocks are stopped. This reduces the wake-up time when external crystals or resonators are used.



### 11.3.5 Extended Standby Mode

Extended Standby mode is identical to Power-save mode with the exception that all enabled system clock sources are kept running while the CPU and Peripheral clocks are stopped. This reduces the wake-up time when external crystals or resonators are used.



# 12. System Control and Reset

#### 12.1 Features

- · Multiple reset sources for safe operation and device reset
  - Power-On Reset
  - External Reset
  - Watchdog Reset

The Watchdog Timer runs from separate, dedicated oscillator

- Brown-Out Reset
  - Accurate, programmable Brown-Out levels
- PDI reset
- Software reset
- · Asynchronous reset
  - No running clock in the device is required for reset
- Reset status register

### 12.2 Resetting the AVR

During reset, all I/O registers are set to their initial values. The SRAM content is not reset. Application execution starts from the Reset Vector. The instruction placed at the Reset Vector should be an Absolute Jump (JMP) instruction to the reset handling routine. By default the Reset Vector address is the lowest Flash program memory address, '0', but it is possible to move the Reset Vector to the first address in the Boot Section.

The I/O ports of the AVR are immediately tri-stated when a reset source goes active.

The reset functionality is asynchronous, so no running clock is required to reset the device.

After the device is reset, the reset source can be determined by the application by reading the Reset Status Register.

### 12.3 Reset Sources

#### 12.3.1 Power-On Reset

The MCU is reset when the supply voltage VCC is below the Power-on Reset threshold voltage.

#### 12.3.2 External Reset

The MCU is reset when a low level is present on the RESET pin.

### 12.3.3 Watchdog Reset

The MCU is reset when the Watchdog Timer period expires and the Watchdog Reset is enabled. The Watchdog Timer runs from a dedicated oscillator independent of the System Clock. For more details see "WDT - Watchdog Timer" on page 24.

#### 12.3.4 Brown-Out Reset

The MCU is reset when the supply voltage VCC is below the Brown-Out Reset threshold voltage and the Brown-out Detector is enabled. The Brown-out threshold voltage is programmable.



#### 12.3.5 PDI reset

The MCU can be reset through the Program and Debug Interface (PDI).

#### 12.3.6 Software reset

The MCU can be reset by the CPU writing to a special I/O register through a timed sequence.

# 13. WDT - Watchdog Timer

#### 13.1 Features

- 11 selectable timeout periods, from 8 ms to 8s.
- Two operation modes
  - Standard mode
  - Window mode
- Runs from the 1 kHz output of the 32 kHz Ultra Low Power oscillator
- Configuration lock to prevent unwanted changes

#### 13.2 Overview

The XMEGA A4 has a Watchdog Timer (WDT). The WDT will run continuously when turned on and if the Watchdog Timer is not reset within a software configurable time-out period, the microcontroller will be reset. The Watchdog Reset (WDR) instruction must be run by software to reset the WDT, and prevent microcontroller reset.

The WDT has a Window mode. In this mode the WDR instruction must be run within a specified period called a window. Application software can set the minimum and maximum limits for this window. If the WDR instruction is not executed inside the window limits, the microcontroller will be reset.

A protection mechanism using a timed write sequence is implemented in order to prevent unwanted enabling, disabling or change of WDT settings.

For maximum safety, the WDT also has an Always-on mode. This mode is enabled by programming a fuse. In Always-on mode, application software can not disable the WDT.



# 14. PMIC - Programmable Multi-level Interrupt Controller

#### 14.1 Features

- Separate interrupt vector for each interrupt
- Short, predictable interrupt response time
- Programmable Multi-level Interrupt Controller
  - 3 programmable interrupt levels
  - Selectable priority scheme within low level interrupts (round-robin or fixed)
  - Non-Maskable Interrupts (NMI)
- Interrupt vectors can be moved to the start of the Boot Section

#### 14.2 Overview

XMEGA A4 has a Programmable Multi-level Interrupt Controller (PMIC). All peripherals can define three different priority levels for interrupts; high, medium or low. Medium level interrupts may interrupt low level interrupt service routines. High level interrupts may interrupt both low-and medium level interrupt service routines. Low level interrupts have an optional round robin scheme to make sure all interrupts are serviced within a certain amount of time.

The built in oscillator failure detection mechanism can issue a Non-Maskable Interrupt (NMI).

# 14.3 Interrupt vectors

When an interrupt is serviced, the program counter will jump to the interrupt vector address. The interrupt vector is the sum of the peripheral's base interrupt address and the offset address for specific interrupts in each peripheral. The base addresses for the XMEGA A4 devices are shown in Table 14-1. Offset addresses for each interrupt available in the peripheral are described for each peripheral in the XMEGA A manual. For peripherals or modules that have only one interrupt, the interrupt vector is shown in Table 14-1. The program address is the word address.

Table 14-1. Reset and Interrupt Vectors

Program Address (Base Address)	Source	Interrupt Description
0x000	RESET	
0x002	OSCF_INT_vect	Crystal Oscillator Failure Interrupt vector (NMI)
0x004	PORTC_INT_base	Port C Interrupt base
0x008	PORTR_INT_base	Port R Interrupt base
0x00C	DMA_INT_base	DMA Controller Interrupt base
0x014	RTC_INT_base	Real Time Counter Interrupt base
0x018	TWIC_INT_base	Two-Wire Interface on Port C Interrupt base
0x01C	TCC0_INT_base	Timer/Counter 0 on port C Interrupt base
0x028	TCC1_INT_base	Timer/Counter 1 on port C Interrupt base
0x030	SPIC_INT_vect	SPI on port C Interrupt vector
0x032	USARTC0_INT_base	USART 0 on port C Interrupt base
0x038	USARTC1_INT_base	USART 1 on port C Interrupt base
0x03E	AES_INT_vect	AES Interrupt vector



 Table 14-1.
 Reset and Interrupt Vectors (Continued)

Program Address (Base Address)	Source	Interrupt Description
0x040	NVM_INT_base	Non-Volatile Memory Interrupt base
0x044	PORTB_INT_base	Port B Interrupt base
0x056	PORTE_INT_base	Port E Interrupt base
0x05A	TWIE_INT_base	Two-Wire Interface on Port E Interrupt base
0x05E	TCE0_INT_base	Timer/Counter 0 on port E Interrupt base
0x06A	TCE1_INT_base	Timer/Counter 1 on port E Interrupt base
0x074	USARTE0_INT_base	USART 0 on port E Interrupt base
0x080	PORTD_INT_base	Port D Interrupt base
0x084	PORTA_INT_base	Port A Interrupt base
0x088	ACA_INT_base	Analog Comparator on Port A Interrupt base
0x08E	ADCA_INT_base	Analog to Digital Converter on Port A Interrupt base
0x09A	TCD0_INT_base	Timer/Counter 0 on port D Interrupt base
0x0A6	TCD1_INT_base	Timer/Counter 1 on port D Interrupt base
0x0AE	SPID_INT_vector	SPI on port D Interrupt vector
0x0B0	USARTD0_INT_base	USART 0 on port D Interrupt base
0x0B6	USARTD1_INT_base	USART 1 on port D Interrupt base



### 15. I/O Ports

#### 15.1 Features

- · Selectable input and output configuration for each pin individually
- Flexible pin configuration through dedicated Pin Configuration Register
- · Synchronous and/or asynchronous input sensing with port interrupts and events
  - Sense both edges
  - Sense rising edges
  - Sense falling edges
  - Sense low level
- · Asynchronous wake-up from all input sensing configurations
- · Two port interrupts with flexible pin masking
- . Highly configurable output driver and pull settings:
  - Totem-pole
  - Pull-up/-down
  - Wired-AND
  - Wired-OR
  - Bus-keeper
  - Inverted I/O
- Optional Slew rate control
- . Configuration of multiple pins in a single operation
- Read-Modify-Write (RMW) support
- Toggle/clear/set registers for Output and Direction registers
- Clock output on port pin
- Event Channel 7 output on port pin
- . Mapping of port registers (virtual ports) into bit accessible I/O memory space

### 15.2 Overview

The XMEGA A4 devices have flexible General Purpose I/O Ports. A port consists of up to 8 pins, ranging from pin 0 to pin 7. The ports implement several functions, including synchronous/asynchronous input sensing, pin change interrupts and configurable output settings. All functions are individual per pin, but several pins may be configured in a single operation.

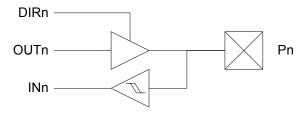
### 15.3 I/O configuration

All port pins (Pn) have programmable output configuration. In addition, all port pins have an inverted I/O function. For an input, this means inverting the signal between the port pin and the pin register. For an output, this means inverting the output signal between the port register and the port pin. The inverted I/O function can be used also when the pin is used for alternate functions.



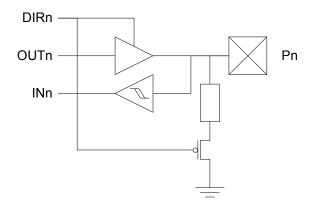
### 15.3.1 Push-pull

Figure 15-1. I/O configuration - Totem-pole



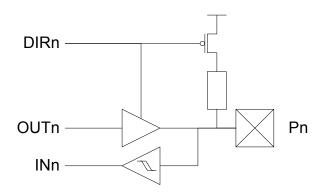
### 15.3.2 Pull-down

Figure 15-2. I/O configuration - Totem-pole with pull-down (on input)



### 15.3.3 Pull-up

Figure 15-3. I/O configuration - Totem-pole with pull-up (on input)

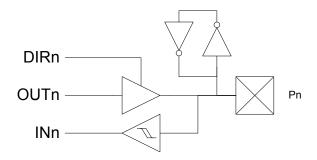


### 15.3.4 Bus-keeper

The bus-keeper's weak output produces the same logical level as the last output level. It acts as a pull-up if the last level was '1', and pull-down if the last level was '0'.



Figure 15-4. I/O configuration - Totem-pole with bus-keeper



### 15.3.5 Others

Figure 15-5. Output configuration - Wired-OR with optional pull-down

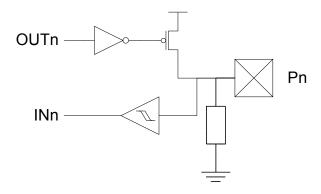
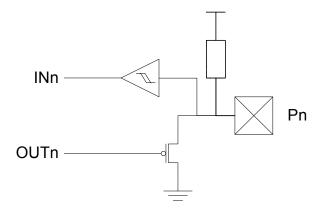


Figure 15-6. I/O configuration - Wired-AND with optional pull-up

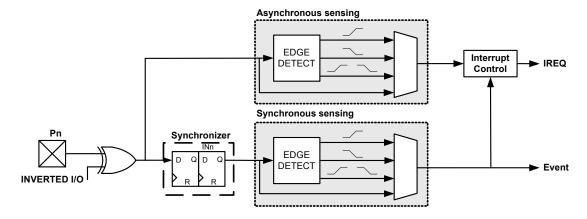


### 15.4 Input sensing

- · Sense both edges
- Sense rising edges
- Sense falling edges
- Sense low level

Input sensing is synchronous or asynchronous depending on the enabled clock for the ports, and the configuration is shown in Figure 15-7 on page 30.

Figure 15-7. Input sensing system overview



When a pin is configured with inverted I/O, the pin value is inverted before the input sensing.

### 15.5 Port Interrupt

Each port has two interrupts with separate priority and interrupt vector. All pins on the port can be individually selected as source for each of the interrupts. The interrupts are then triggered according to the input sense configuration for each pin configured as source for the interrupt.

### 15.6 Alternate Port Functions

In addition to the input/output functions on all port pins, most pins have alternate functions. This means that other modules or peripherals connected to the port can use the port pins for their functions, such as communication or pulse-width modulation. "Pinout and Pin Functions" on page 49 shows which modules on peripherals that enable alternate functions on a pin, and which alternate function is available on a pin.



### 16. T/C - 16-bit Timer/Counter

#### 16.1 Features

- Five 16-bit Timer/Counters
  - Three Timer/Counters of type 0
  - Two Timer/Counters of type 1
- Three Compare or Capture (CC) Channels in Timer/Counter 0
- Two Compare or Capture (CC) Channels in Timer/Counter 1
- Double Buffered Timer Period Setting
- Double Buffered Compare or Capture Channels
- Waveform Generation:
  - Single Slope Pulse Width Modulation
  - Dual Slope Pulse Width Modulation
  - Frequency Generation
- Input Capture:
  - Input Capture with Noise Cancelling
  - Frequency capture
  - Pulse width capture
  - 32-bit input capture
- Event Counter with Direction Control
- Timer Overflow and Timer Error Interrupts and Events
- One Compare Match or Capture Interrupt and Event per CC Channel
- Supports DMA Operation
- Hi-Resolution Extension (Hi-Res)
- Advanced Waveform Extension (AWEX)

#### 16.2 Overview

XMEGA A4 has five Timer/Counters, three Timer/Counter 0 and two Timer/Counter 1. The difference between them is that Timer/Counter 0 has four Compare/Capture channels, while Timer/Counter 1 has two Compare/Capture channels.

The Timer/Counters (T/C) are 16-bit and can count any clock, event or external input in the microcontroller. A programmable prescaler is available to get a useful T/C resolution. Updates of Timer and Compare registers are double buffered to ensure glitch free operation. Single slope PWM, dual slope PWM and frequency generation waveforms can be generated using the Compare Channels.

Through the Event System, any input pin or event in the microcontroller can be used to trigger input capture, hence no dedicated pins are required for this. The input capture has a noise canceller to avoid incorrect capture of the T/C, and can be used to do frequency and pulse width measurements.

A wide range of interrupt or event sources are available, including T/C Overflow, Compare match and Capture for each Compare/Capture channel in the T/C.

PORTC and PORTD each has one Timer/Counter 0 and one Timer/Counter1. PORTE has one Timer/Conter0. Notation of these are TCC0 (Time/Counter C0), TCC1, TCD0, TCD1 and TCE0, respectively.



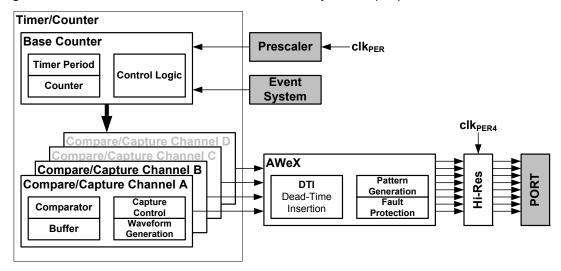


Figure 16-1. Overview of a Timer/Counter and closely related peripherals

The Hi-Resolution Extension can be enabled to increase the waveform generation resolution by 2 bits (4x). This is available for all Timer/Counters. See "Hi-Res - High Resolution Extension" on page 34 for more details.

The Advanced Waveform Extension can be enabled to provide extra and more advanced feature for the Timer/Counter. This is only available for Timer/Counter 0. See "AWEX - Advanced Waveform Extension" on page 33 for more details.



### 17. AWEX - Advanced Waveform Extension

### 17.1 Features

- Output with complementary output from each Capture channel
- Four Dead Time Insertion (DTI) Units, one for each Capture channel
- 8-bit DTI Resolution
- Separate High and Low Side Dead-Time Setting
- Double Buffered Dead-Time
- Event Controlled Fault Protection
- Single Channel Multiple Output Operation (for BLDC motor control)
- Double Buffered Pattern Generation

### 17.2 Overview

The Advanced Waveform Extension (AWEX) provides extra features to the Timer/Counter in Waveform Generation (WG) modes. The AWEX enables easy and safe implementation of for example, advanced motor control (AC, BLDC, SR, and Stepper) and power control applications.

Any WG output from a Timer/Counter 0 is split into a complimentary pair of outputs when any AWEX feature is enabled. These output pairs go through a Dead-Time Insertion (DTI) unit that enables generation of the non-inverted Low Side (LS) and inverted High Side (HS) of the WG output with dead time insertion between LS and HS switching. The DTI output will override the normal port value according to the port override setting. Optionally the final output can be inverted by using the invert I/O setting for the port pin.

The Pattern Generation unit can be used to generate a synchronized bit pattern on the port it is connected to. In addition, the waveform generator output from Compare Channel A can be distributed to, and override all port pins. When the Pattern Generator unit is enabled, the DTI unit is bypassed.

The Fault Protection unit is connected to the Event System. This enables any event to trigger a fault condition that will disable the AWEX output. Several event channels can be used to trigger fault on several different conditions.

The AWEX is available for TCC0. The notation of this is AWEXC.



# 18. Hi-Res - High Resolution Extension

### 18.1 Features

- Increases Waveform Generator resolution by 2-bits (4x)
- Supports Frequency, single- and dual-slope PWM operation
- Supports the AWEX when this is enabled and used for the same Timer/Counter

### 18.2 Overview

The Hi-Resolution (Hi-Res) Extension is able to increase the resolution of the waveform generation output by a factor of 4. When enabled for a Timer/Counter, the Fast Peripheral clock running at four times the CPU clock speed will be as input to the Timer/Counter.

The High Resolution Extension can also be used when an AWEX is enabled and used with a Timer/Counter.

XMEGA A4 devices have three Hi-Res Extensions that each can be enabled for each Timer/Counters pair on PORTC, PORTD and PORTE. The notation of these are HIRESC, HIRESD and HIRESE, respectively.



### 19. RTC - 16-bit Real-Time Counter

#### 19.1 Features

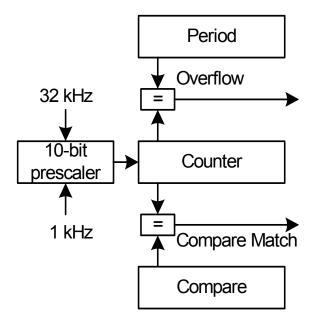
- 16-bit Timer
- Flexible Tick resolution ranging from 1 Hz to 32.768 kHz
- One Compare register
- One Period register
- Clear timer on Overflow or Compare Match
- Overflow or Compare Match event and interrupt generation

#### 19.2 Overview

The XMEGA A4 includes a 16-bit Real-time Counter (RTC). The RTC can be clocked from an accurate 32.768 kHz Crystal Oscillator, the 32.768 kHz Calibrated Internal Oscillator, or from the 32 kHz Ultra Low Power Internal Oscillator. The RTC includes both a Period and a Compare register. For details, see Figure 19-1.

A wide range of Resolution and Time-out periods can be configured using the RTC. With a maximum resolution of 30.5  $\mu$ s, time-out periods range up to 2000 seconds. With a resolution of 1 second, the maximum time-out period is over 18 hours (65536 seconds).

Figure 19-1. Real Time Counter overview





### 20. TWI - Two-Wire Interface

#### 20.1 Features

- Two Identical TWI peripherals
- Simple yet Powerful and Flexible Communication Interface
- Both Master and Slave Operation Supported
- Device can Operate as Transmitter or Receiver
- 7-bit Address Space Allows up to 128 Different Slave Addresses
- Multi-master Arbitration Support
- Up to 400 kHz Data Transfer Speed
- Slew-rate Limited Output Drivers
- Noise Suppression Circuitry Rejects Spikes on Bus Lines
- Fully Programmable Slave Address with General Call Support
- Address Recognition Causes Wake-up when in Sleep Mode
- I<sup>2</sup>C and System Management Bus (SMBus) compatible

### 20.2 Overview

The Two-Wire Interface (TWI) is a bi-directional wired-AND bus with only two lines, the clock (SCL) line and the data (SDA) line. The protocol makes it possible to interconnect up to 128 individually addressable devices. Since it is a multi-master bus, one or more devices capable of taking control of the bus can be connected.

The only external hardware needed to implement the bus is a single pull-up resistor for each of the TWI bus lines. Mechanisms for resolving bus contention are inherent in the TWI protocol.

PORTC and PORTE each has one TWI. Notation of these peripherals are TWIC and TWIE, respectively.



# 21. SPI - Serial Peripheral Interface

### 21.1 Features

- Two Identical SPI peripherals
- Full-duplex, Three-wire Synchronous Data Transfer
- Master or Slave Operation
- LSB First or MSB First Data Transfer
- Seven Programmable Bit Rates
- End of Transmission Interrupt Flag
- Write Collision Flag Protection
- Wake-up from Idle Mode
- Double Speed (CK/2) Master SPI Mode

## 21.2 Overview

The Serial Peripheral Interface (SPI) allows high-speed full-duplex, synchronous data transfer between different devices. Devices can communicate using a master-slave scheme, and data is transferred both to and from the devices simultaneously.

PORTC and PORTD each has one SPI. Notation of these peripherals are SPIC and SPID, respectively.



## 22. USART

#### 22.1 Features

- Five Identical USART peripherals
- Full Duplex Operation (Independent Serial Receive and Transmit Registers)
- Asynchronous or Synchronous Operation
- Master or Slave Clocked Synchronous Operation
- High-resolution Arithmetic Baud Rate Generator
- Supports Serial Frames with 5, 6, 7, 8, or 9 Data Bits and 1 or 2 Stop Bits
- Odd or Even Parity Generation and Parity Check Supported by Hardware
- Data OverRun Detection
- Framing Error Detection
- . Noise Filtering Includes False Start Bit Detection and Digital Low Pass Filter
- Three Separate Interrupts on TX Complete, TX Data Register Empty and RX Complete
- Multi-processor Communication Mode
- Double Speed Asynchronous Communication Mode
- . Master SPI mode for SPI communication
- IrDA support through the IRCOM module

## 22.2 Overview

The Universal Synchronous and Asynchronous serial Receiver and Transmitter (USART) is a highly flexible serial communication module. The USART supports full duplex communication, and both asynchronous and clocked synchronous operation. The USART can also be set in Master SPI mode to be used for SPI communication.

Communication is frame based, and the frame format can be customized to support a wide range of standards. The USART is buffered in both direction, enabling continued data transmission without any delay between frames. There are separate interrupt vectors for receive and transmit complete, enabling fully interrupt driven communication. Frame error and buffer overflow are detected in hardware and indicated with separate status flags. Even or odd parity generation and parity check can also be enabled.

One USART can use the IRCOM module to support IrDA 1.4 physical compliant pulse modulation and demodulation for baud rates up to 115.2 kbps.

PORTC and PORTD each has two USARTs. PORTE has one USART. Notation of these peripherals are USARTC0, USARTC1, USARTD0, USARTD1 and USARTE0, respectively.



## 23. IRCOM - IR Communication Module

### 23.1 Features

- Pulse modulation/demodulation for infrared communication
- Compatible to IrDA 1.4 physical for baud rates up to 115.2 kbps
- Selectable pulse modulation scheme
  - 3/16 of baud rate period
  - Fixed pulse period, 8-bit programmable
  - Pulse modulation disabled
- · Built in filtering
- · Can be connected to and used by one USART at a time

#### 23.2 Overview

XMEGA contains an Infrared Communication Module (IRCOM) for IrDA communication with baud rates up to 115.2 kbps. This supports three modulation schemes: 3/16 of baud rate period, fixed programmable pulse time based on the Peripheral Clock speed, or pulse modulation disabled. There is one IRCOM available which can be connected to any USART to enable infrared pulse coding/decoding for that USART.



## 24. Crypto Engine

#### 24.1 Features

- Data Encryption Standard (DES) CPU instruction
- Advanced Encryption Standard (AES) Crypto module
- DES Instruction
  - Encryption and Decryption
  - Single-cycle DES instruction
  - Encryption/Decryption in 16 clock cycles per 8-byte block
- AES Crypto Module
  - Encryption and Decryption
  - Support 128-bit keys
  - Support XOR data load mode to the State memory for Cipher Block Chaining
  - Encryption/Decryption in 375 clock cycles per 16-byte block

#### 24.2 Overview

The Advanced Encryption Standard (AES) and Data Encryption Standard (DES) are two commonly used encryption standards. These are supported through an AES peripheral module and a DES CPU instruction. All communication interfaces and the CPU can optionally use AES and DES encrypted communication and data storage.

DES is supported by a DES instruction in the AVR XMEGA CPU. The 8-byte key and 8-byte data blocks must be loaded into the Register file, and then DES must be executed 16 times to encrypt/decrypt the data block.

The AES Crypto Module encrypts and decrypts 128-bit data blocks with the use of a 128-bit key. The key and data must be loaded into the key and state memory in the module before encryption/decryption is started. It takes 375 peripheral clock cycles before the encryption/decryption is done and decrypted/encrypted data can be read out, and an optional interrupt can be generated. The AES Crypto Module also has DMA support with transfer triggers when encryption/decryption is done and optional auto-start of encryption/decryption when the state memory is fully loaded.



## 25. ADC - 12-bit Analog to Digital Converter

#### 25.1 Features

- One ADC with 12-bit resolution
- · 2 Msps sample rate
- Signed and Unsigned conversions
- · 4 result registers with individual input channel control
- 12 single ended inputs
- 8x4 differential inputs
- · 4 internal inputs:
  - Integrated Temperature Sensor
  - DAC Output
  - VCC voltage divided by 10
  - Bandgap voltage
- Software selectable gain of 2, 4, 8, 16, 32 or 64
- Selectable accuracy of 8- or 12-bit.
- Internal or External Reference selection
- · Event triggered conversion for accurate timing
- DMA transfer of conversion results
- Interrupt/Event on compare result

#### 25.2 Overview

XMEGA A4 devices have one Analog to Digital Converter (ADC), see Figure 25-1 on page 42.

The ADC converts analog voltages to digital values. The ADC has 12-bit resolution and is capable of converting up to 2 million samples per second. The input selection is flexible, and both single-ended and differential measurements can be done. For differential measurements an optional gain stage is available to increase the dynamic range. In addition several internal signal inputs are available. The ADC can provide both signed and unsigned results.

This is a pipeline ADC. A pipeline ADC consists of several consecutive stages, where each stage convert one part of the result. The pipeline design enables high sample rate at low clock speeds, and remove limitations on samples speed versus propagation delay. This also means that a new analog voltage can be sampled and a new ADC measurement started while other ADC measurements are ongoing.

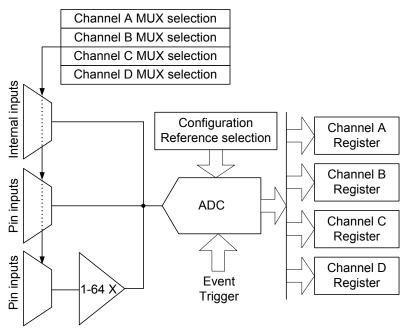
ADC measurements can either be started by application software or an incoming event from another peripheral in the device. Four different result registers with individual input selection (MUX selection) are provided to make it easier for the application to keep track of the data. Each result register and MUX selection pair is referred to as an ADC Channel. It is possible to use DMA to move ADC results directly to memory or peripherals when conversions are done.

Both internal and external analog reference voltages can be used. An accurate internal 1.0V reference is available.

An integrated temperature sensor is available and the output from this can be measured with the ADC. The output from the DAC, VCC/10 and the Bandgap voltage can also be measured by the ADC.



Figure 25-1. ADC overview



Each ADC has four MUX selection registers with a corresponding result register. This means that four channels can be sampled within 1.5 µs without any intervention by the application other than starting the conversion. The results will be available in the result registers.

The ADC may be configured for 8- or 12-bit resolution, reducing the minimum conversion time (propagation delay) from  $3.5 \,\mu s$  for 12-bit to  $2.5 \,\mu s$  for 8-bit resolution.

ADC conversion results are provided left- or right adjusted with optional '1' or '0' padding. This eases calculation when the result is represented as a signed integer (signed 16-bit number).

PORTA has one ADC. Notation of this peripheral is ADCA.



## 26. DAC - 12-bit Digital to Analog Converter

### 26.1 Features

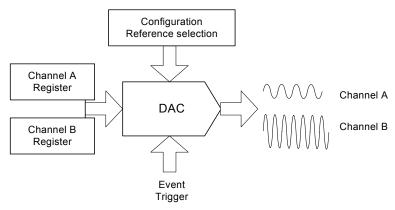
- · One DAC with 12-bit resolution
- . Up to 1 Msps conversion rate
- Flexible conversion range
- Multiple trigger sources
- 1 continuous output or 2 Sample and Hold (S/H) outputs
- · Built-in offset and gain calibration
- · High drive capabilities
- Low Power Mode

#### 26.2 Overview

The XMEGA A4 devices feature one 12-bit, 1 Msps DAC with built-in offset and gain calibration, see Figure 26-1 on page 43.

A DAC converts a digital value into an analog signal. The DAC may use an internal 1.1 voltage as the upper limit for conversion, but it is also possible to use the supply voltage or any applied voltage in-between. The external reference input is shared with the ADC reference input.

Figure 26-1. DAC overview



The DAC has one continuous output with high drive capabilities for both resistive and capacitive loads. It is also possible to split the continuous time channel into two Sample and Hold (S/H) channels, each with separate data conversion registers.

A DAC conversion may be started from the application software by writing the data conversion registers. The DAC can also be configured to do conversions triggered by the Event System to have regular timing, independent of the application software. DMA may be used for transferring data from memory locations to DAC data registers.

The DAC has a built-in calibration system to reduce offset and gain error when loading with a calibration value from software.

PORTB has one DAC. Notation of this peripheral is DACB.



# 27. AC - Analog Comparator

#### 27.1 Features

- Two Analog Comparators
- Selectable Power vs. Speed
- Selectable hysteresis
  - 0, 20 mV, 50 mV
- Analog Comparator output available on pin
- Flexible Input Selection
  - All pins on the port
  - Output from the DAC
  - Bandgap reference voltage.
  - Voltage scaler that can perform a 64-level scaling of the internal VCC voltage.
- Interrupt and event generation on
  - Rising edge
  - Falling edge
  - Toggle
- · Window function interrupt and event generation on
  - Signal above window
  - Signal inside window
  - Signal below window

#### 27.2 Overview

XMEGA A4 features two Analog Comparators (AC). An Analog Comparator compares two voltages, and the output indicates which input is largest. The Analog Comparator may be configured to give interrupt requests and/or events upon several different combinations of input change.

Both hysteresis and propagation delays may be adjusted in order to find the optimal operation for each application.

A wide range of input selection is available, both external pins and several internal signals can be used.

The Analog Comparators are always grouped in pairs (AC0 and AC1) on each analog port. They have identical behavior but separate control registers.

Optionally, the state of the comparator is directly available on a pin.

PORTA has one AC pair. Notation of this peripheral is ACA.



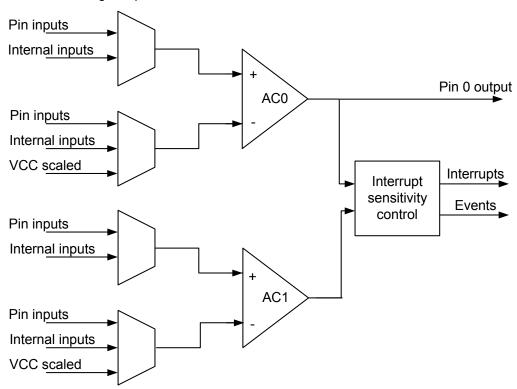


Figure 27-1. Analog comparator overview



## 27.3 Input Selection

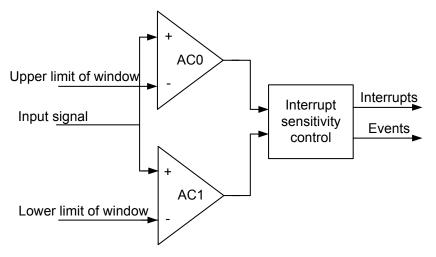
The Analog comparators have a very flexible input selection and the two comparators grouped in a pair may be used to realize a window function. One pair of analog comparators is shown in Figure 27-1 on page 45.

- Input selection from pin
  - Pin 0, 1, 2, 3, 4, 5, 6 selectable to positive input of analog comparator
  - Pin 0, 1, 3, 5, 7 selectable to negative input of analog comparator
- Internal signals available on positive analog comparator inputs
  - Output from 12-bit DAC
- Internal signals available on negative analog comparator inputs
  - 64-level scaler of the VCC, available on negative analog comparator input
  - Bandgap voltage reference
  - Output from 12-bit DAC

### 27.4 Window Function

The window function is realized by connecting the external inputs of the two analog comparators in a pair as shown in Figure 27-2.

Figure 27-2. Analog comparator window function





# 28. OCD - On-chip Debug

#### 28.1 Features

- Complete Program Flow Control
  - Go, Stop, Reset, Step into, Step over, Step out, Run-to-Cursor
- Debugging on C and high-level language source code level
- Debugging on Assembler and disassembler level
- · 1 dedicated program address or source level breakpoint for AVR Studio / debugger
- 4 Hardware Breakpoints
- Unlimited Number of User Program Breakpoints
- Unlimited Number of User Data Breakpoints, with break on:
  - Data location read, write or both read and write
  - Data location content equal or not equal to a value
  - Data location content is greater or less than a value
  - Data location content is within or outside a range
  - Bits of a data location are equal or not equal to a value
- Non-Intrusive Operation
  - No hardware or software resources in the device are used
- High Speed Operation
  - No limitation on debug/programming clock frequency versus system clock frequency

#### 28.2 Overview

The XMEGA A4 has a powerful On-Chip Debug (OCD) system that - in combination with Atmel's development tools - provides all the necessary functions to debug an application. It has support for program and data breakpoints, and can debug an application from C and high level language source code level, as well as assembler and disassembler level. It has full Non-Intrusive Operation and no hardware or software resources in the device are used. The ODC system is accessed through an external debugging tool which connects to the PDI physical interface. Refer to "Program and Debug Interfaces" on page 48.



# 29. Program and Debug Interfaces

### 29.1 Features

- PDI Program and Debug Interface (Atmel proprietary 2-pin interface)
- Access to the OCD system
- Programming of Flash, EEPROM, Fuses and Lock Bits

#### 29.2 Overview

The programming and debug facilities are accessed through PDI physical interface. The PDI physical interface uses one dedicated pin together with the Reset pin, and no general purpose pins are used.

# 29.3 PDI - Program and Debug Interface

The PDI is an Atmel proprietary protocol for communication between the microcontroller and Atmel's development tools.



## 30. Pinout and Pin Functions

The pinout of XMEGA A4 is shown in "Pinout/Block Diagram" on page 3. In addition to general I/O functionality, each pin may have several functions. This will depend on which peripheral is enabled and connected to the actual pin. Only one of the alternate pin functions can be used at time.

## 30.1 Alternate Pin Functions Description

The tables below shows the notation for all pin functions available and describe their functions.

### 30.1.1 Operation/Power Supply

VCC Digital supply voltage
AVCC Analog supply voltage

GND Ground

### 30.1.2 Port Interrupt functions

SYNC Port pin with full synchronous and limited asynchronous interrupt function
ASYNC Port pin with full synchronous and full asynchronous interrupt function

#### 30.1.3 Analog functions

ACn Analog Comparator input pin n
AC0OUT Analog Comparator 0 Output

ADCn Analog to Digital Converter input pin n

DACn Digital to Analog Converter output pin n

AREF Analog Reference input pin

### 30.1.4 Timer/Counter and AWEX functions

OCnx Output Compare Channel x for Timer/Counter n

OCnx Inverted Output Compare Channel x for Timer/Counter n



### 30.1.5 Communication functions

SCL Serial Clock for TWI
SDA Serial Data for TWI

XCKn Transfer Clock for USART n

RXDn Receiver Data for USART n

TXDn Transmitter Data for USART n

SIave Select for SPI

MOSI Master Out Slave In for SPI
MISO Master In Slave Out for SPI

SCK Serial Clock for SPI

## 30.1.6 Oscillators, Clock and Event

TOSCn Timer Oscillator pin n

XTALn Input/Output for inverting Oscillator pin n

## 30.1.7 Debug/System functions

RESET Reset pin

PDI\_CLK Program and Debug Interface Clock pin
PDI\_DATA Program and Debug Interface Data pin



## 30.2 Alternate Pin Functions

The tables below shows the main and alternate pin functions for all pins on each port. It also shows which peripheral which make use of or enable the alternate pin function.

Table 30-1. Port A - Alternate functions

PORTA	PIN#	INTERRUPT	ADCA POS	ADCA NEG	ADCA GAINPOS	ADCA GAINNEG	ACA POS	ACA NEG	ACA OUT	REF
GND	38									
AVCC	39									
PA0	40	SYNC	ADC0	ADC0	ADC0		AC0	AC0		AREF
PA1	41	SYNC	ADC1	ADC1	ADC1		AC1	AC1		
PA2	42	SYNC/ASYNC	ADC2	ADC2	ADC2		AC2			
PA3	43	SYNC	ADC3	ADC3	ADC3		AC3	AC3		
PA4	44	SYNC	ADC4		ADC4	ADC4	AC4			
PA5	1	SYNC	ADC5		ADC5	ADC5	AC5	AC5		
PA6	2	SYNC	ADC6		ADC6	ADC6	AC6			
PA7	3	SYNC	ADC7		ADC7	ADC7		AC7	AC0 OUT	

Table 30-2. Port B - Alternate functions

PORTB	PIN#	INTERRUPT	ADCA POS	DACB	REF
PB0	4	SYNC	ADC8		AREF
PB1	5	SYNC	ADC9		
PB2	6	SYNC/ASYNC	ADC10	DAC0	
PB3	7	SYNC	ADC11	DAC1	

Table 30-3. Port C - Alternate functions

PORTC	PIN#	INTERRUPT	TCC0	AWEXC	TCC1	USARTC0	USARTC1	SPI	TWIC	CLOCKOUT	EVENTOUT
GND	8										
vcc	9										
PC0	10	SYNC	OC0A	OC0A					SDA		
PC1	11	SYNC	OC0B	OC0A		XCK0			SCL		
PC2	12	SYNC/ASYNC	OC0C	OC0B		RXD0					
PC3	13	SYNC	OC0D	OC0B		TXD0					
PC4	14	SYNC		OC0C	OC1A			SS			
PC5	15	SYNC		OC0C	OC1B		XCK1	MOSI			
PC6	16	SYNC		OC0D			RXD1	MISO			
PC7	17	SYNC		OC0D			TXD1	SCK		CLKOUT	EVOUT



Table 30-4. Port D - Alternate functions

PORTD	PIN#	INTERRUPT	TCD0	TCD1	USARTD0	USARTD1	SPID	CLOCKOUT	EVENTOUT
GND	18								
vcc	19								
PD0	20	SYNC	OC0A						
PD1	21	SYNC	OC0B		XCK0				
PD2	22	SYNC/ASYNC	OC0C		RXD0				
PD3	23	SYNC	OC0D		TXD0				
PD4	24	SYNC		OC1A			SS		
PD5	25	SYNC		OC1B		XCK1	MOSI		
PD6	26	SYNC				RXD1	MISO		
PD7	27	SYNC				TXD1	SCK	CLKOUT	EVOUT

**Table 30-5.** Port E - Alternate functions

PORT E	PIN#	INTERRUPT	TCE0	USARTE0	TWIE
PE0	28	SYNC	OC0A		SDA
PE1	29	SYNC	OC0B	хско	SCL
GND	30				
vcc	31				
PE2	32	SYNC/ASYNC	OC0C	RXD0	
PE3	33	SYNC	OCOD	TXD0	

**Table 30-6.** Port R - Alternate functions

PORTR	PIN#	XTAL	PDI	тоѕс
PDI	34		PDI_DATA	
RESET	35		PDI_CLK	
PR0	36	XTAL2		TOSC2
PR1	37	XTAL1		TOSC1



# 31. Peripheral Module Address Map

The address maps show the base address for each peripheral and module in XMEGA A4. For complete register description and summary for each peripheral module, refer to the XMEGA A Manual.

Base Address	Name	Description
0x0000	GPIO	General Purpose IO Registers
0x0010	VPORT0	Virtual Port 0
0x0014	VPORT1	Virtual Port 1
0x0018	VPORT2	Virtual Port 2
0x001C	VPORT3	Virtual Port 2
0x0030	CPU	CPU
0x0040	CLK	Clock Control
0x0048	SLEEP	Sleep Controller
0x0050	OSC	Oscillator Control
0x0060	DFLLRC32M	DFLL for the 32 MHz Internal RC Oscillator
0x0068	DFLLRC2M	DFLL for the 2 MHz RC Oscillator
0x0070	PR	Power Reduction
0x0078	RST	Reset Controller
0x0080	WDT	Watch-Dog Timer
0x0090	MCU	MCU Control
0x00A0	PMIC	Programmable MUltilevel Interrupt Controller
0x00B0	PORTCFG	Port Configuration
0x00C0	AES	AES Module
0x0100	DMA	DMA Controller
0x0180	EVSYS	Event System
0x01C0	NVM	Non Volatile Memory (NVM) Controller
0x0200	ADCA	Analog to Digital Converter on port A
0x0320	DACB	Digital to Analog Converter on port B
0x0380	ACA	Analog Comparator pair on port A
0x0400	RTC	Real Time Counter
0x0480	TWIC	Two Wire Interface on port C
0x04A0	TWIE	Two Wire Interface on port E
0x0600	PORTA	Port A
0x0620	PORTB	Port B
0x0640	PORTC	Port C
0x0660	PORTD	Port D
0x0680	PORTE	Port E
0x07E0	PORTR	Port R
0x0800	TCC0	Timer/Counter 0 on port C
0x0840	TCC1	Timer/Counter 1 on port C
0x0880	AWEXC	Advanced Waveform Extension on port C
0x0890	HIRESC	High Resolution Extension on port C
0x08A0	USARTC0	USART 0 on port C
0x08B0	USARTC1	USART 1 on port C
0x08C0	SPIC	Serial Peripheral Interface on port C
0x08F8	IRCOM	Infrared Communication Module
0x0900	TCD0	Timer/Counter 0 on port D
0x0940	TCD1	Timer/Counter 1 on port D
0x0990	HIRESD	High Resolution Extension on port D
0x09A0	USARTD0	USART 0 on port D
0x09B0	USARTD1	USART 1 on port D
0x09C0	SPID	Serial Peripheral Interface on port D
0x0A00	TCE0	Timer/Counter 0 on port E
0x0A90	HIRESE	High Resolution Extension on port E
0x0AA0	USARTE0	USART 0 on port E



# 32. Instruction Set Summary

Mnemonics	Operands	Description	Opera	ation		Flags	#Clocks
	'	Arithmetic	and Logic Instructions			'	
ADD	Rd, Rr	Add without Carry	Rd	<b>←</b>	Rd + Rr	Z,C,N,V,S,H	1
ADC	Rd, Rr	Add with Carry	Rd	<b>←</b>	Rd + Rr + C	Z,C,N,V,S,H	1
ADIW	Rd, K	Add Immediate to Word	Rd	<b>←</b>	Rd + 1:Rd + K	Z,C,N,V,S	2
SUB	Rd, Rr	Subtract without Carry	Rd	<b>←</b>	Rd - Rr	Z,C,N,V,S,H	1
SUBI	Rd, K	Subtract Immediate	Rd	<b>←</b>	Rd - K	Z,C,N,V,S,H	1
SBC	Rd, Rr	Subtract with Carry	Rd	<b>←</b>	Rd - Rr - C	Z,C,N,V,S,H	1
SBCI	Rd, K	Subtract Immediate with Carry	Rd	<b>←</b>	Rd - K - C	Z,C,N,V,S,H	1
SBIW	Rd, K	Subtract Immediate from Word	Rd + 1:Rd	<b>←</b>	Rd + 1:Rd - K	Z,C,N,V,S	2
AND	Rd, Rr	Logical AND	Rd	<b>←</b>	Rd • Rr	Z,N,V,S	1
ANDI	Rd, K	Logical AND with Immediate	Rd	<b>←</b>	Rd • K	Z,N,V,S	1
OR	Rd, Rr	Logical OR	Rd	<b>←</b>	Rd v Rr	Z,N,V,S	1
ORI	Rd, K	Logical OR with Immediate	Rd	<b>←</b>	Rd v K	Z,N,V,S	1
EOR	Rd, Rr	Exclusive OR	Rd	<b>←</b>	Rd ⊕ Rr	Z,N,V,S	1
СОМ	Rd	One's Complement	Rd	<b>←</b>	\$FF - Rd	Z,C,N,V,S	1
NEG	Rd	Two's Complement	Rd	<b>←</b>	\$00 - Rd	Z,C,N,V,S,H	1
SBR	Rd,K	Set Bit(s) in Register	Rd	<b>←</b>	Rd v K	Z,N,V,S	1
CBR	Rd,K	Clear Bit(s) in Register	Rd	<b>←</b>	Rd • (\$FFh - K)	Z,N,V,S	1
INC	Rd	Increment	Rd	<b>←</b>	Rd + 1	Z,N,V,S	1
DEC	Rd	Decrement	Rd	<b>←</b>	Rd - 1	Z,N,V,S	1
TST	Rd	Test for Zero or Minus	Rd	<b>←</b>	Rd • Rd	Z,N,V,S	1
CLR	Rd	Clear Register	Rd	<b>←</b>	Rd ⊕ Rd	Z,N,V,S	1
SER	Rd	Set Register	Rd	<b>←</b>	\$FF	None	1
MUL	Rd,Rr	Multiply Unsigned	R1:R0	<b>←</b>	Rd x Rr (UU)	Z,C	2
MULS	Rd,Rr	Multiply Signed	R1:R0	<b>←</b>	Rd x Rr (SS)	Z,C	2
MULSU	Rd,Rr	Multiply Signed with Unsigned	R1:R0	<b>←</b>	Rd x Rr (SU)	Z,C	2
FMUL	Rd,Rr	Fractional Multiply Unsigned	R1:R0	<b>←</b>	Rd x Rr<<1 (UU)	Z,C	2
FMULS	Rd,Rr	Fractional Multiply Signed	R1:R0	$\leftarrow$	Rd x Rr<<1 (SS)	Z,C	2
FMULSU	Rd,Rr	Fractional Multiply Signed with Unsigned	R1:R0	←	Rd x Rr<<1 (SU)	Z,C	2
DES	К	Data Encryption	if (H = 0) then R15:R0 else if (H = 1) then R15:R0	<b>←</b>	Encrypt(R15:R0, K) Decrypt(R15:R0, K)		1/2
		Bra	nch Instructions			1	
RJMP	k	Relative Jump	PC	<b>←</b>	PC + k + 1	None	2
IJMP		Indirect Jump to (Z)	PC(15:0) PC(21:16)	<b>←</b>	Z, 0	None	2
EIJMP		Extended Indirect Jump to (Z)	PC(15:0) PC(21:16)	<b>←</b>	Z, EIND	None	2
JMP	k	Jump	PC	<b>←</b>	k	None	3
RCALL	k	Relative Call Subroutine	PC	<b>←</b>	PC + k + 1	None	2/3 <sup>(1)</sup>
ICALL		Indirect Call to (Z)	PC(15:0) PC(21:16)	<b>←</b>	Z, 0	None	2/3 <sup>(1)</sup>
EICALL		Extended Indirect Call to (Z)	PC(15:0) PC(21:16)	<b>←</b>	Z, EIND	None	3 <sup>(1)</sup>



Mnemonics	Operands	Description	Opera	ation		Flags	#Clocks
CALL	k	call Subroutine	PC	<b>←</b>	k	None	3 / 4 <sup>(1)</sup>
RET		Subroutine Return	PC	<b>←</b>	STACK	None	4 / 5 <sup>(1)</sup>
RETI		Interrupt Return	PC	<b>←</b>	STACK	I	4 / 5 <sup>(1)</sup>
CPSE	Rd,Rr	Compare, Skip if Equal	if (Rd = Rr) PC	<b>←</b>	PC + 2 or 3	None	1/2/3
СР	Rd,Rr	Compare	Rd - Rr			Z,C,N,V,S,H	1
CPC	Rd,Rr	Compare with Carry	Rd - Rr - C			Z,C,N,V,S,H	1
CPI	Rd,K	Compare with Immediate	Rd - K			Z,C,N,V,S,H	1
SBRC	Rr, b	Skip if Bit in Register Cleared	if (Rr(b) = 0) PC	<b>←</b>	PC + 2 or 3	None	1/2/3
SBRS	Rr, b	Skip if Bit in Register Set	if (Rr(b) = 1) PC	<b>←</b>	PC + 2 or 3	None	1/2/3
SBIC	A, b	Skip if Bit in I/O Register Cleared	if (I/O(A,b) = 0) PC	<b>←</b>	PC + 2 or 3	None	2/3/4
SBIS	A, b	Skip if Bit in I/O Register Set	If (I/O(A,b) =1) PC	<b>←</b>	PC + 2 or 3	None	2/3/4
BRBS	s, k	Branch if Status Flag Set	if (SREG(s) = 1) then PC	<b>←</b>	PC + k + 1	None	1/2
BRBC	s, k	Branch if Status Flag Cleared	if (SREG(s) = 0) then PC	<b>←</b>	PC + k + 1	None	1/2
BREQ	k	Branch if Equal	if (Z = 1) then PC	<b>←</b>	PC + k + 1	None	1/2
BRNE	k	Branch if Not Equal	if (Z = 0) then PC	<b>←</b>	PC + k + 1	None	1/2
BRCS	k	Branch if Carry Set	if (C = 1) then PC	<b>←</b>	PC + k + 1	None	1/2
BRCC	k	Branch if Carry Cleared	if (C = 0) then PC	<b>←</b>	PC + k + 1	None	1/2
BRSH	k	Branch if Same or Higher	if (C = 0) then PC	<b>←</b>	PC + k + 1	None	1/2
BRLO	k	Branch if Lower	if (C = 1) then PC	<b>←</b>	PC + k + 1	None	1/2
BRMI	k	Branch if Minus	if (N = 1) then PC	<b>←</b>	PC + k + 1	None	1/2
BRPL	k	Branch if Plus	if (N = 0) then PC	<b>←</b>	PC + k + 1	None	1/2
BRGE	k	Branch if Greater or Equal, Signed	if (N ⊕ V= 0) then PC	<b>←</b>	PC + k + 1	None	1/2
BRLT	k	Branch if Less Than, Signed	if (N ⊕ V= 1) then PC	<b>←</b>	PC + k + 1	None	1/2
BRHS	k	Branch if Half Carry Flag Set	if (H = 1) then PC	<b>←</b>	PC + k + 1	None	1/2
BRHC	k	Branch if Half Carry Flag Cleared	if (H = 0) then PC	<b>←</b>	PC + k + 1	None	1/2
BRTS	k	Branch if T Flag Set	if (T = 1) then PC	<b>←</b>	PC + k + 1	None	1/2
BRTC	k	Branch if T Flag Cleared	if (T = 0) then PC	<b>←</b>	PC + k + 1	None	1/2
BRVS	k	Branch if Overflow Flag is Set	if (V = 1) then PC	<b>←</b>	PC + k + 1	None	1/2
BRVC	k	Branch if Overflow Flag is Cleared	if (V = 0) then PC	<b>←</b>	PC + k + 1	None	1/2
BRIE	k	Branch if Interrupt Enabled	if (I = 1) then PC	<b>←</b>	PC + k + 1	None	1/2
BRID	k	Branch if Interrupt Disabled	if (I = 0) then PC	<b>←</b>	PC + k + 1	None	1/2
	+	Data T	ransfer Instructions				
MOV	Rd, Rr	Copy Register	Rd	<b>←</b>	Rr	None	1
MOVW	Rd, Rr	Copy Register Pair	Rd+1:Rd	<b>←</b>	Rr+1:Rr	None	1
LDI	Rd, K	Load Immediate	Rd	<b>←</b>	К	None	1
LDS	Rd, k	Load Direct from data space	Rd	<b>←</b>	(k)	None	2 <sup>(1)(2)</sup>
LD	Rd, X	Load Indirect	Rd	<b>←</b>	(X)	None	1(1)(2)
LD	Rd, X+	Load Indirect and Post-Increment	Rd X	<b>←</b>	(X) X + 1	None	1 <sup>(1)(2)</sup>
LD	Rd, -X	Load Indirect and Pre-Decrement	$X \leftarrow X - 1$ , $Rd \leftarrow (X)$	<b>←</b>	X - 1 (X)	None	2 <sup>(1)(2)</sup>
LD	Rd, Y	Load Indirect	$Rd \leftarrow (Y)$	<b>←</b>	(Y)	None	1(1)(2)
LD	Rd, Y+	Load Indirect and Post-Increment	Rd Y	<b>←</b>	(Y) Y + 1	None	1 <sup>(1)(2)</sup>



Mnemonics	Operands	Description	Opera	ation		Flags	#Clocks
LD	Rd, -Y	Load Indirect and Pre-Decrement	Y Rd	<b>←</b>	Y - 1 (Y)	None	2 <sup>(1)(2)</sup>
LDD	Rd, Y+q	Load Indirect with Displacement	Rd	<b>←</b>	(Y + q)	None	2 <sup>(1)(2)</sup>
LD	Rd, Z	Load Indirect	Rd	<b>←</b>	(Z)	None	1 <sup>(1)(2)</sup>
LD	Rd, Z+	Load Indirect and Post-Increment	Rd Z	<b>←</b>	(Z), Z+1	None	1(1)(2)
LD	Rd, -Z	Load Indirect and Pre-Decrement	Z Rd	<b>←</b>	Z - 1, (Z)	None	2 <sup>(1)(2)</sup>
LDD	Rd, Z+q	Load Indirect with Displacement	Rd	<b>←</b>	(Z + q)	None	2 <sup>(1)(2)</sup>
STS	k, Rr	Store Direct to Data Space	(k)	←	Rd	None	2 <sup>(1)</sup>
ST	X, Rr	Store Indirect	(X)	<b>←</b>	Rr	None	1 <sup>(1)</sup>
ST	X+, Rr	Store Indirect and Post-Increment	(X) X	<b>←</b>	Rr, X + 1	None	1 <sup>(1)</sup>
ST	-X, Rr	Store Indirect and Pre-Decrement	X (X)	<b>←</b> <b>←</b>	X - 1, Rr	None	2 <sup>(1)</sup>
ST	Y, Rr	Store Indirect	(Y)	<b>←</b>	Rr	None	1 <sup>(1)</sup>
ST	Y+, Rr	Store Indirect and Post-Increment	(Y) Y	<b>←</b>	Rr, Y + 1	None	1 <sup>(1)</sup>
ST	-Y, Rr	Store Indirect and Pre-Decrement	Y (Y)	<b>←</b>	Y - 1, Rr	None	2 <sup>(1)</sup>
STD	Y+q, Rr	Store Indirect with Displacement	(Y + q)	<b>←</b>	Rr	None	2 <sup>(1)</sup>
ST	Z, Rr	Store Indirect	(Z)	<b>←</b>	Rr	None	1 <sup>(1)</sup>
ST	Z+, Rr	Store Indirect and Post-Increment	(Z) Z	<b>←</b>	Rr Z + 1	None	1 <sup>(1)</sup>
ST	-Z, Rr	Store Indirect and Pre-Decrement	Z	<b>←</b>	Z - 1	None	2 <sup>(1)</sup>
STD	Z+q,Rr	Store Indirect with Displacement	(Z + q)	<b>←</b>	Rr	None	2 <sup>(1)</sup>
LPM		Load Program Memory	R0	<b>←</b>	(Z)	None	3
LPM	Rd, Z	Load Program Memory	Rd	←	(Z)	None	3
LPM	Rd, Z+	Load Program Memory and Post-Increment	Rd Z	<b>←</b>	(Z), Z + 1	None	3
ELPM		Extended Load Program Memory	R0	<b>←</b>	(RAMPZ:Z)	None	3
ELPM	Rd, Z	Extended Load Program Memory	Rd	←	(RAMPZ:Z)	None	3
ELPM	Rd, Z+	Extended Load Program Memory and Post- Increment	Rd Z	<b>←</b>	(RAMPZ:Z), Z + 1	None	3
SPM		Store Program Memory	(RAMPZ:Z)	<b>←</b>	R1:R0	None	-
SPM	Z+	Store Program Memory and Post-Increment by 2	(RAMPZ:Z) Z	<b>←</b>	R1:R0, Z + 2	None	-
IN	Rd, A	In From I/O Location	Rd	<b>←</b>	I/O(A)	None	1
OUT	A, Rr	Out To I/O Location	I/O(A)	<b>←</b>	Rr	None	1
PUSH	Rr	Push Register on Stack	STACK	<b>←</b>	Rr	None	1 <sup>(1)</sup>
POP	Rd	Pop Register from Stack	Rd	<b>←</b>	STACK	None	2 <sup>(1)</sup>
		Bit and	Bit-test Instructions				
LSL	Rd	Logical Shift Left	Rd(n+1) Rd(0) C	<b>← ← ←</b>	Rd(n), 0, Rd(7)	Z,C,N,V,H	1
LSR	Rd	Logical Shift Right	Rd(n) Rd(7) C	← ← ←	Rd(n+1), 0, Rd(0)	Z,C,N,V	1



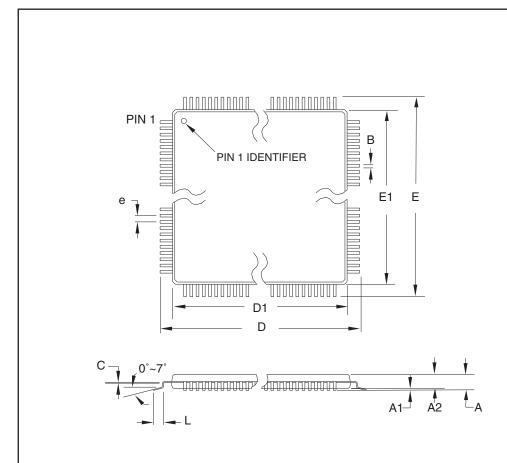
Mnemonics	Operands	Description	Oper	ation		Flags	#Clocks
ROL	Rd	Rotate Left Through Carry	Rd(0) Rd(n+1) C	← ← ←	C, Rd(n), Rd(7)	Z,C,N,V,H	1
ROR	Rd	Rotate Right Through Carry	Rd(7) Rd(n) C	← ← ←	C, Rd(n+1), Rd(0)	Z,C,N,V	1
ASR	Rd	Arithmetic Shift Right	Rd(n)	<b>←</b>	Rd(n+1), n=06	Z,C,N,V	1
SWAP	Rd	Swap Nibbles	Rd(30)	$\leftrightarrow$	Rd(74)	None	1
BSET	s	Flag Set	SREG(s)	<b>←</b>	1	SREG(s)	1
BCLR	s	Flag Clear	SREG(s)	<b>←</b>	0	SREG(s)	1
SBI	A, b	Set Bit in I/O Register	I/O(A, b)	<b>←</b>	1	None	1
СВІ	A, b	Clear Bit in I/O Register	I/O(A, b)	<b>←</b>	0	None	1
BST	Rr, b	Bit Store from Register to T	Т	<b>←</b>	Rr(b)	Т	1
BLD	Rd, b	Bit load from T to Register	Rd(b)	<b>←</b>	Т	None	1
SEC		Set Carry	С	<b>←</b>	1	С	1
CLC		Clear Carry	С	<b>←</b>	0	С	1
SEN		Set Negative Flag	N	<b>←</b>	1	N	1
CLN		Clear Negative Flag	N	<b>←</b>	0	N	1
SEZ		Set Zero Flag	Z	<b>←</b>	1	Z	1
CLZ		Clear Zero Flag	Z	<b>←</b>	0	Z	1
SEI		Global Interrupt Enable	I	<b>←</b>	1	1	1
CLI		Global Interrupt Disable	1	<b>←</b>	0	I	1
SES		Set Signed Test Flag	S	<b>←</b>	1	S	1
CLS		Clear Signed Test Flag	S	<b>←</b>	0	S	1
SEV		Set Two's Complement Overflow	V	<b>←</b>	1	V	1
CLV		Clear Two's Complement Overflow	V	<b>←</b>	0	V	1
SET		Set T in SREG	Т	<b>←</b>	1	Т	1
CLT		Clear T in SREG	Т	<b>←</b>	0	Т	1
SEH		Set Half Carry Flag in SREG	Н	<b>←</b>	1	Н	1
CLH		Clear Half Carry Flag in SREG	Н	<b>←</b>	0	Н	1
	•	мси	Control Instructions				•
BREAK		Break	(See specific de	scr. fo	r BREAK)	None	1
NOP		No Operation				None	1
SLEEP		Sleep	(see specific de	escr. fo	or Sleep)	None	1
WDR		Watchdog Reset	(see specific d	escr. fo	or WDR)	None	1

- Notes: 1. Cycle times for Data memory accesses assume internal memory accesses, and are not valid for accesses via the external RAM interface.
  - 2. One extra cycle must be added when accessing Internal SRAM.



# 33. Packaging information

## 33.1 44A



## **COMMON DIMENSIONS**

(Unit of Measure = mm)

SYMBOL	MIN	NOM	MAX	NOTE			
Α	_	_	1.20				
A1	0.05	_	0.15				
A2	0.95	1.00	1.05				
D	11.75	12.00	12.25				
D1	9.90	10.00	10.10	Note 2			
Е	11.75	12.00	12.25				
E1	9.90	10.00	10.10	Note 2			
В	0.30	-	0.45				
С	0.09	_	0.20				
L	0.45	_	0.75				
е		0.80 TYP					

10/5/2001

Notes:

- 1. This package conforms to JEDEC reference MS-026, Variation ACB.
- Dimensions D1 and E1 do not include mold protrusion. Allowable protrusion is 0.25 mm per side. Dimensions D1 and E1 are maximum plastic body size dimensions including mold mismatch.
- 3. Lead coplanarity is 0.10 mm maximum.

AMEL 232 Sar

2325 Orchard Parkway San Jose, CA 95131

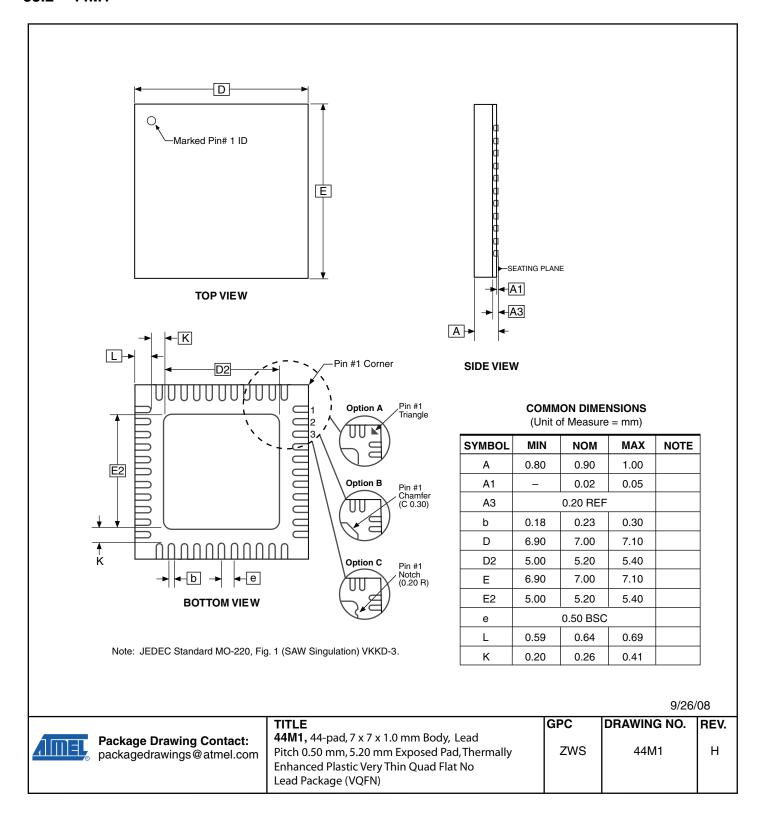
#### TITLE

**44A**, 44-lead, 10 x 10 mm Body Size, 1.0 mm Body Thickness, 0.8 mm Lead Pitch, Thin Profile Plastic Quad Flat Package (TQFP)

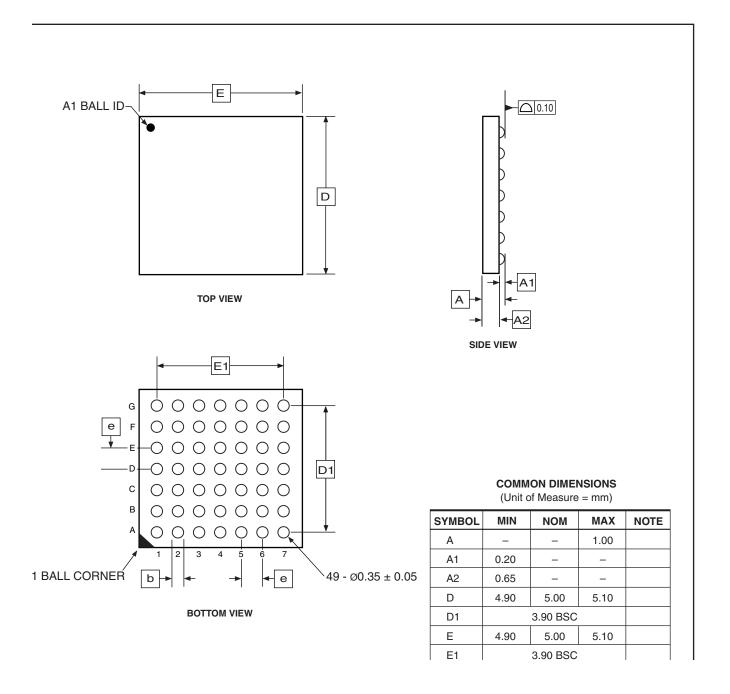
DRAWING NO.	REV.
44A	В



### 33.2 44M1



## 33.3 49C2





## 34. Electrical Characteristics

# 34.1 Absolute Maximum Ratings\*

Operating Temperature55°C to +125°C
Storage Temperature65°C to +150°C
Voltage on any Pin with respect to Ground0.5V to V <sub>CC</sub> +0.5V
Maximum Operating Voltage
DC Current per I/O Pin
DC Current V <sub>CC</sub> and GND Pins200.0 mA

\*NOTICE:

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or other conditions beyond those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## 34.2 DC Characteristics

#### **Current Consumption**

Symbol	Parameter	Conditi	ion		Min.	Тур.	Max.	Units
			20 kHz Evt Olk	V <sub>CC</sub> = 1.8V		30		
			32 kHz, Ext. Clk	V <sub>CC</sub> = 3.0V		75		
			1 MHz, Ext. Clk	V <sub>CC</sub> = 1.8V		260		μA
		Active		V <sub>CC</sub> = 3.0V		570		
			OMIL For Oil	V <sub>CC</sub> = 1.8V		510	690	
			2 MHz, Ext. Clk	V <sub>CC</sub> = 3.0V		1.1	1.49	
	Davier Complex Comment(1)		32 MHz, Ext. Clk	V <sub>CC</sub> = 3.0V		11.4	13	mA
	Power Supply Current <sup>(1)</sup>		00 LLL - F.+ OIL	V <sub>CC</sub> = 1.8V		2.8		
			32 kHz, Ext. Clk	V <sub>CC</sub> = 3.0V		4.8		
		Idle	1 MHz, Ext. Clk	V <sub>CC</sub> = 1.8V		80		μΑ
				V <sub>CC</sub> = 3.0V		150		
I <sub>cc</sub>			2 MHz, Ext. Clk	V <sub>CC</sub> = 1.8V		160	225	
I ICC				V <sub>CC</sub> = 3.0V		295	390	
			32 MHz, Ext. Clk	V <sub>CC</sub> = 3.0V		4.8	6	mA
		All Fund	ctions Disabled	V <sub>CC</sub> = 3.0V		0.1		
		All Fund	ctions Disabled, T = 85°C	V <sub>CC</sub> = 3.0V		1.5	5	
	Power-down mode	шьм	DT Compled DOD	V <sub>CC</sub> = 1.8V		1.1		
		ULP, W	DT, Sampled BOD	V <sub>CC</sub> = 3.0V		1.1		μΑ
		ULP, W	DT, Sampled BOD, T=85°C	V <sub>CC</sub> = 3.0V		2.6	10	
		RTC 1 I	kHz from Low Power 32 kHz	V <sub>CC</sub> = 1.8V		0.52		
	Power-save mode	TOSC				0.61		
		RTC fro	RTC from Low Power 32 kHz TOSC $V_{CC} = 3$ .			1.16		
	Reset Current Consumption	without	Reset pull-up resistor current	V <sub>CC</sub> = 3.0V		505		



## **Current Consumption (Continued)**

Symbol	Parameter	Condition	Min.	Тур.	Max.	Units	
Module o	current consumption <sup>(2)</sup>						
	RC32M			470			
	RC32M w/DFLL	Internal 32.768 kHz oscillator as DFLL source		600			
	RC2M			112			
	RC2M w/DFLL	Internal 32.768 kHz oscillator as DFLL source		145			
	RC32K			30			
	PLL	Multiplication factor = 10x		225			
	Watchdog normal mode			0.9		μΑ	
	BOD Continuous mode			120		μ",	
	BOD Sampled mode			1			
	Internal 1.00 V ref			80			
	Temperature reference			80			
I <sub>CC</sub>	RTC with int. 32 kHz RC as source	No prescaling		30			
	RTC with ULP as source	No prescaling		0.9			
	ADC	250 kS/s - Int. 1V Ref		2.9			
	DAC Normal Mode	Single channel, Int. 1V Ref		2.4		mA	
	DAC Low-Power Mode	Single channel, Int. 1V Ref		1.1			
	AC High-speed			280			
	AC Low-power			110			
	USART	Rx and Tx enabled, 9600 BAUD		5.3			
	DMA			95		μΑ	
	Timer/Counter	Prescaler DIV1		19			
	AES			140			

Note:

- 1. All Power Reduction Registers set. T = 25°C if not specified.
- 2. All parameters measured as the difference in current consumption between module enabled and disabled. All data at  $V_{CC} = 3.0V$ ,  $Clk_{SYS} = 1$  MHz External clock with no prescaling,  $T = 25^{\circ}C$ .



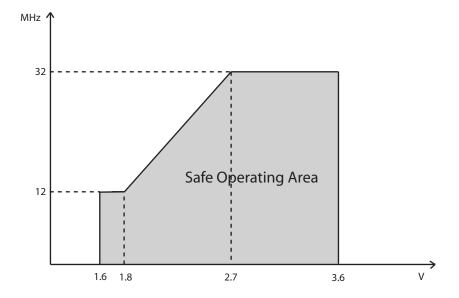
# 34.3 Speed

Table 34-1. Speed

Symbol	Parameter	Condition	Min	Тур	Max	Units
		V <sub>CC</sub> = 1.6V	0		12	
System clock	System clock	V <sub>CC</sub> = 1.8V	0		12	N 41 1-
Clk <sub>SYS</sub>	frequency	V <sub>CC</sub> = 2.7V	0		32	MHz
		V <sub>CC</sub> = 3.6V	0		32	

The maximum System clock frequency of the XMEGA A4 devices is depending on  $V_{CC}$ . As shown in Figure 34-1 on page 63 the Frequency vs.  $V_{CC}$  curve is linear between  $1.8V < V_{CC} < 2.7V$ .

Figure 34-1. Operating Frequency vs.Vcc





#### Flash and EEPROM Memory Characteristics 34.4

Table 34-2. **Endurance and Data Retention** 

Symbol	Parameter	Condition		Min	Тур	Max	Units	
	Flash	Weite/France avales	25°C	10K			Cuala	
		Write/Erase cycles	85°C	10K			Cycle	
			Data waterstiers	25°C	100			V
		Data retention	55°C	25			- Year	
			25°C	80K				
	EEPROM -	Write/Erase cycles	85°C	30K			Cycle	
			25°C	100				
		Data retention	55°C	25			Year	

Table 34-3. Programming time

Symbol	Parameter	Condition	Min	Typ <sup>(1)</sup>	Max	Units
	Chip Erase	Flash, EEPROM <sup>(2)</sup> and SRAM Erase		40		
Flas		Page Erase		6		
	Flash	Page Write		6		
		Page WriteAutomatic Page Erase and Write		12		ms
		Page Erase		6		
	EEPROM	Page Write		6		
		Page WriteAutomatic Page Erase and Write		12		

- Notes: 1. Programming is timed from the internal 2 MHz oscillator.
  - 2. EEPROM is not erased if the EESAVE fuse is programmed.

## 34.5 ADC Characteristics

Table 34-4. ADC Characteristics

Symbol	Parameter	Condition	Min	Тур	Max	Units
RES	Resolution	Programmable: 8/12	8	12	12	Bits
INL	Integral Non-Linearity	500 ksps	-5	±2	5	LSB
DNL	Differential Non-Linearity	500 ksps		< ±1		LSB
	Gain Error			< ±10		mV
	Offset Error			< ±2		mV
ADC <sub>clk</sub>	ADC Clock frequency	Max is 1/4 of Peripheral Clock			2000	kHz
	Conversion rate				2000	ksps
	Conversion time (propagation delay)	(RES+2)/2+GAIN RES = 8 or 12, GAIN = 0 or 1	5	7	8	ADC <sub>clk</sub> cycles
	Sampling Time	1/2 ADC <sub>clk</sub> cycle	0.25			uS
	Conversion range		0		VREF	V
VREF	Reference voltage		1.0		V <sub>cc</sub> -0.6V	V
	Input bandwidth					kHz
INT1V	Internal 1.00V reference			1.00		V
INTVCC	Internal V <sub>CC</sub> /1.6			V <sub>CC</sub> /1.6		V
SCALEDVCC	Scaled internal V <sub>CC</sub> /10 input			V <sub>CC</sub> /10		V
R <sub>AREF</sub>	Reference input resistance			> 10		MΩ
	Start-up time					μs
	Internal input sampling speed	Temp. sensor, V <sub>CC</sub> /10, Bandgap			100	ksps

 Table 34-5.
 ADC Gain Stage Characteristics

Symbol	Parameter	Coi	Min	Тур	Max	Units	
	Gain error	1 to 64 gain			< ±1		%
	Offset error				< ±1		
\/w	Naise level at innut	C.A. crain	VREF = Int. 1V		0.12		mV
Vrms	Noise level at input	64x gain  VREF = Ext. 2V			0.06		
	Clock rate	Same as ADC				1000	kHz



## 34.6 DAC Characteristics

Table 34-6. DAC Characteristics

Symbol	Parameter	Cor	ndition	Min	Тур	Max	Units
INL	Integral Non-Linearity	$V_{CC} = 1.6-3.6V$	VREF = Ext. ref		±5		
DNL	Differential New Linearity	V 1000V	VREF = Ext. ref		<±1		LSB
DINL DI	Differential Non-Linearity	$V_{CC} = 1.6-3.6V$	VREF= AV <sub>CC</sub>				
F <sub>clk</sub>	Conversion rate					1000	ksps
AREF	External reference voltage			1.1		AV <sub>CC</sub> -0.6	٧
	Reference input impedance				>10		MΩ
	DC output impedance						kΩ
	Max output voltage	$R_{load}=100k\Omega$			AV <sub>CC</sub> *0.98		V
	Min output voltage	$R_{load}=100k\Omega$			<0.030		V
	Offset factory calibration accuracy	Continues mode, V	Continues mode, V <sub>CC</sub> =3.0V,		±0.5		LSB
	Gain factory calibration accuracy	VREF = Int 1.00V, 7	=85°C		±2.5		LOD

# 34.7 Analog Comparator Characteristics

Table 34-7. Analog Comparator Characteristics

Symbol	Parameter	Condition		Min	Тур	Max	Units	
V <sub>off</sub>	Input Offset Voltage	V <sub>CC</sub> = 1.6 - 3.6V			< ±10		mV	
I <sub>lk</sub>	Input Leakage Current	$V_{CC} = 1.6 - 3.6V$	V <sub>CC</sub> = 1.6 - 3.6V		< 1000		pA	
V <sub>hys1</sub>	Hysteresis, No	V <sub>CC</sub> = 1.6 - 3.6V			0		mV	
V <sub>hys2</sub>	Hysteresis, Small	V <sub>CC</sub> = 1.6 - 3.6V	mode = HS		20		\/	
V <sub>hys3</sub>	Hysteresis, Large	V <sub>CC</sub> = 1.6 - 3.6V	mode = HS		40		mV	
	Draw a station delay.	V <sub>CC</sub> = 3.0V, T= 85°C	mode = HS		90			
t <sub>delay</sub>	Propagation delay	V <sub>CC</sub> = 1.6 - 3.6V	mode = LP		175		ns	

# 34.8 Bandgap Characteristics

 Table 34-8.
 Bandgap Voltage Characteristics

Symbol	Parameter	Condition	Min	Тур	Max	Units
	Bandson Startun Timo	As reference to ADC or DAC	1 CI	k_PER + 2	.5µs	
	Bandgap Startup Time	As input to AC or ADC		1.5		μs
	Bandgap voltage			1.1		
	ADC/DAC mat	T= 85°C, After calibration	0.99	1	1.01	V
	ADC/DAC ref			1		
	Variation over voltage and temperature	$V_{CC} = 1.6 - 3.6V$ , $T_A = -40$ °C to 85°C		±5		%



## 34.9 Brownout Detection Characteristics

**Table 34-9.** Brownout Detection Characteristics<sup>(1)</sup>

Symbol	Parameter	Condition	Min	Тур	Max	Units
	BOD level 0 falling Vcc			1.62		
	BOD level 1 falling Vcc			1.9		
	BOD level 2 falling Vcc			2.17		
	BOD level 3 falling Vcc			2.43		.,
	BOD level 4 falling Vcc			2.68		V
	BOD level 5 falling Vcc			2.96		
	BOD level 6 falling Vcc			3.22		
	BOD level 7 falling Vcc			3.49		
	Hysteresis	BOD level 0-5		1		%

Note: 1. BOD is calibrated on BOD level 0 at 85°C.

## 34.10 PAD Characteristics

Table 34-10. PAD Characteristics

Symbol	Parameter	Condition	Min	Тур	Max	Units
V	Innet High Voltage	V <sub>CC</sub> = 2.4 - 3.6V	0.7*V <sub>CC</sub>		V <sub>CC</sub> +0.5	
V <sub>IH</sub>	Input High Voltage	V <sub>CC</sub> = 1.6 - 2.4V	0.8*V <sub>CC</sub>		V <sub>CC</sub> +0.5	
V	Input Low Voltage	V <sub>CC</sub> = 2.4 - 3.6V	-0.5		0.3*V <sub>CC</sub>	
V <sub>IL</sub>	Input Low Voltage	V <sub>CC</sub> = 1.6 - 2.4V	-0.5		0.2*V <sub>CC</sub>	
		$I_{OH} = 15 \text{ mA}, V_{CC} = 3.3 \text{V}$		0.4	0.76	V
V <sub>OL</sub>	Output Low Voltage GPIO	I <sub>OH</sub> = 10 mA, V <sub>CC</sub> = 3.0V		0.3	0.64	"
		$I_{OH} = 5 \text{ mA}, V_{CC} = 1.8V$		0.2	0.46	
	Output High Voltage GPIO	$I_{OH} = -8 \text{ mA}, V_{CC} = 3.3 \text{V}$	2.6	3.0		
V <sub>OH</sub>		$I_{OH} = -6 \text{ mA}, V_{CC} = 3.0 \text{V}$	2.1	2.7		
		$I_{OH} = -2 \text{ mA}, V_{CC} = 1.8 \text{V}$	1.4	1.6		
I <sub>IL</sub>	Input Leakage Current I/O pin			<0.001	1	
I <sub>IH</sub>	Input Leakage Current I/O pin			<0.001	1	μA
R <sub>P</sub>	I/O pin Pull/Buss keeper Resistor	T= -40°C to 85°C		20		l <sub>1</sub> O
R <sub>RST</sub>	Reset pin Pull-up Resistor	T= -40°C to 85°C		20		kΩ
	Input hysteresis	V <sub>CC</sub> = 1.6 V - 3.6 V, T= -40°C to 85°C		0.5		V

## 34.11 POR Characteristics

Table 34-11. Power-on Reset Characteristics

Symbol	Parameter	Condition	Min	Тур	Max	Units
V <sub>POT-</sub>	POR threshold voltage falling Vcc			1		.,
V <sub>POT+</sub>	POR threshold voltage rising Vcc			1.3		V

## 34.12 Reset Characteristics

## Table 34-12. Reset Characteristics

Symbol	Parameter	Condition	Min	Тур	Max	Units
	Minimum reset pulse width			90		ns
	Reset threshold voltage	V <sub>CC</sub> = 2.7 - 3.6V		0.45*V <sub>CC</sub>		V
		V <sub>CC</sub> = 1.6 - 2.7V		0.42*V <sub>CC</sub>		V

## 34.13 Oscillator Characteristics

## Table 34-13. Internal 32.768 kHz Oscillator Characteristics

Symbol	Parameter	Condition	Min	Тур	Max	Units
	Accuracy	$T = 85$ °C, $V_{CC} = 3V$ , After production calibration	-0.5		0.5	%

## Table 34-14. Internal 2 MHz Oscillator Characteristics

Symbol	Parameter	Condition	Min	Тур	Max	Units
	Accuracy	$T = 85$ °C, $V_{CC} = 3V$ , After production calibration	-1.5		1.5	%
	DFLL Calibration step size	T = 25°C, V <sub>CC</sub> = 3V		0.15		

#### Table 34-15. Internal 32 MHz Oscillator Characteristics

Symbol	Parameter	Condition	Min	Тур	Max	Units
	Accuracy	$T = 85$ °C, $V_{CC} = 3V$ , After production calibration	-1.5		1.5	%
	DFLL Calibration stepsize	$T = 25^{\circ}C, V_{CC} = 3V$		0.2		

## Table 34-16. Internal 32 kHz, ULP Oscillator Characteristics

Symbol	Parameter	Condition	Min	Тур	Max	Units
	Output frequency 32 kHz ULP OSC	$T = 85^{\circ}C, V_{CC} = 3.0V$		26		kHz



# 35. Typical Characteristics

# 35.1 Active Supply Current

Figure 35-1. Active Supply Current vs. Frequency

 $f_{SYS} = 0$  - 1.0 MHz External clock, T = 25°C.

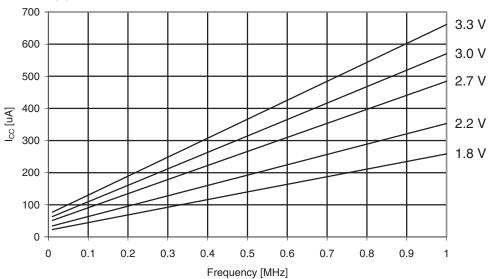


Figure 35-2. Active Supply Current vs. Frequency

 $f_{SYS} = 1 - 32$  MHz External clock, T = 25°C.

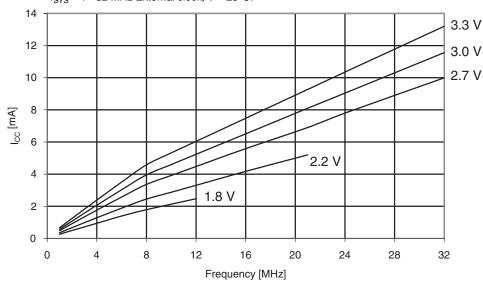




Figure 35-3. Active Supply Current vs. Vcc

 $f_{SYS} = 1.0 \text{ MHz External Clock.}$ 

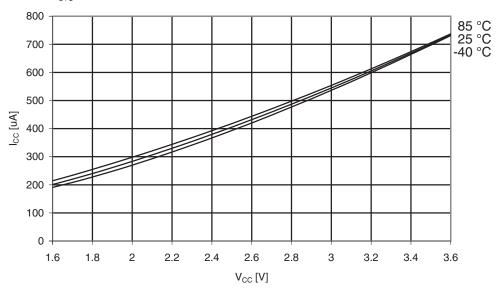


Figure 35-4. Active Supply Current vs. VCC

 $f_{SYS} = 32.768 \text{ kHz internal RC}.$ 

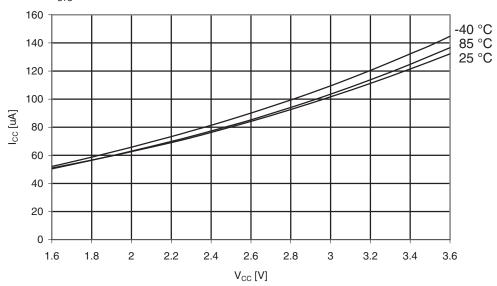


Figure 35-5. Active Supply Current vs. Vcc

 $f_{SYS} = 2.0 \text{ MHz internal RC}.$ 

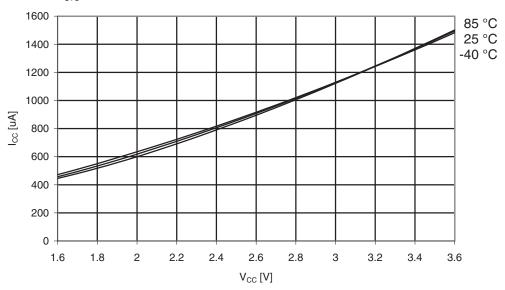
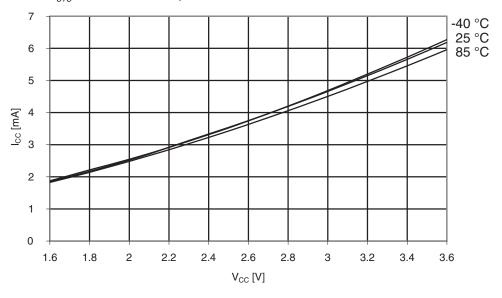


Figure 35-6. Active Supply Current vs. Vcc

 $f_{SYS} = 32$  MHz internal RC prescaled to 8 MHz.

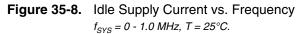




 $f_{SYS} = 32 \text{ MHz internal RC}.$ 16 -40 °C 25 °C 14 85 °C 12 10  $l_{cc}$  [mA] 8 6 4 2 0 2.7 2.8 2.9 3 3.1 3.2 3.3 3.4 3.5 3.6  $V_{CC}[V]$ 

Figure 35-7. Active Supply Current vs. Vcc

# 35.2 Idle Supply Current



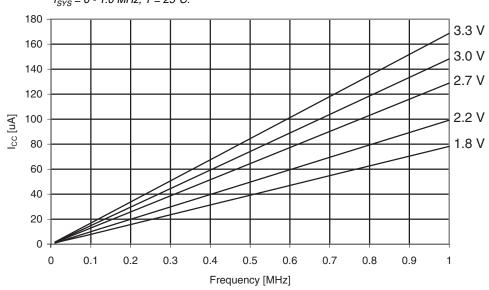


Figure 35-9. Idle Supply Current vs. Frequency

 $f_{SYS}=1$  - 32 MHz,  $T=25^{\circ}C$ .

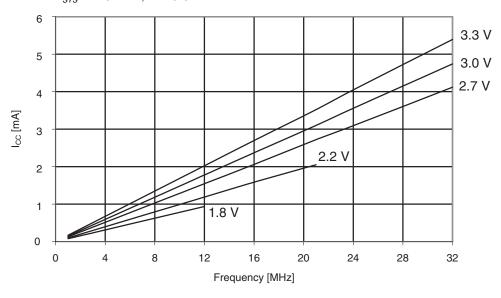


Figure 35-10. Idle Supply Current vs. Vcc

 $f_{SYS} = 1.0 \text{ MHz External Clock.}$ 

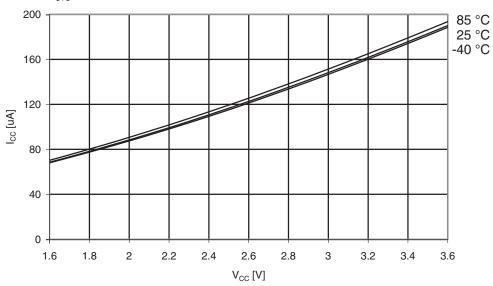




Figure 35-11. Idle Supply Current vs. Vcc

 $f_{SYS} = 32.768 \text{ kHz internal RC}.$ 

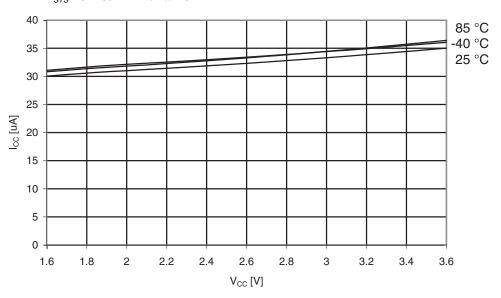


Figure 35-12. Idle Supply Current vs. Vcc

 $f_{SYS} = 2.0 \text{ MHz internal RC}.$ 

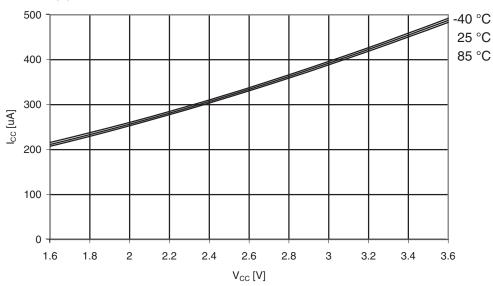


Figure 35-13. Idle Supply Current vs. Vcc

 $f_{SYS} = 32$  MHz internal RC prescaled to 8 MHz.

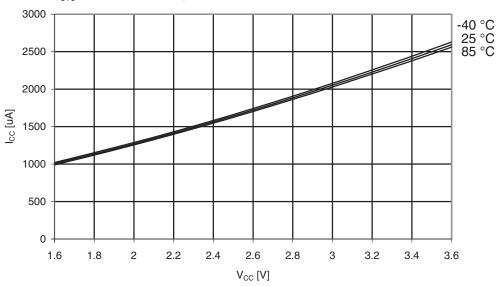
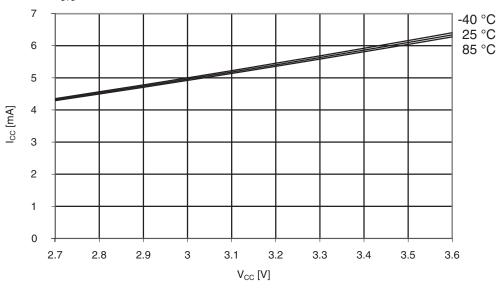


Figure 35-14. Idle Supply Current vs. Vcc

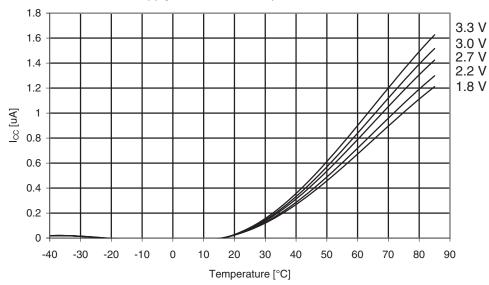
 $f_{SYS} = 32 \text{ MHz internal RC}.$ 



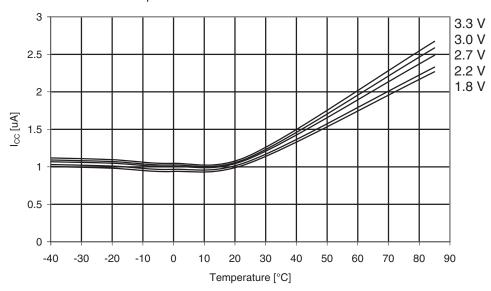


# 35.3 Power-down Supply Current

Figure 35-15. Power-down Supply Current vs. Temperature



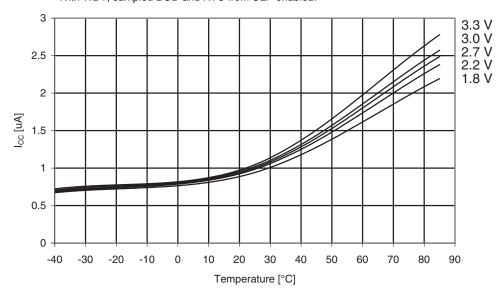
**Figure 35-16.** Power-down Supply Current vs. Temperature With WDT and sampled BOD enabled.





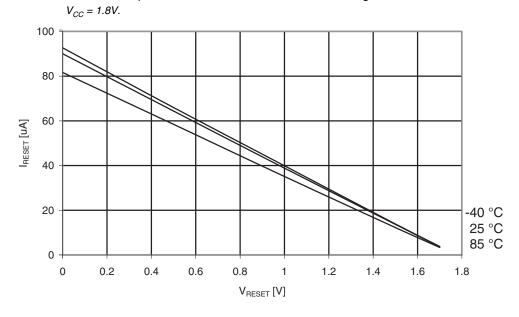
# 35.4 Power-save Supply Current

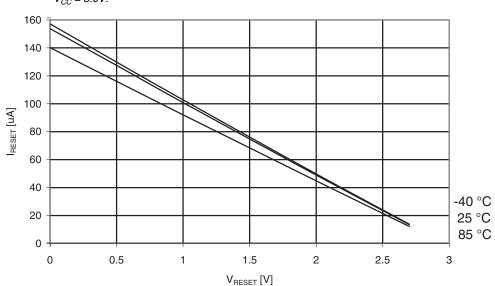
**Figure 35-17.** Power-save Supply Current vs. Temperature With WDT, sampled BOD and RTC from ULP enabled.



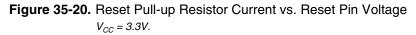
# 35.5 Pin Pull-up

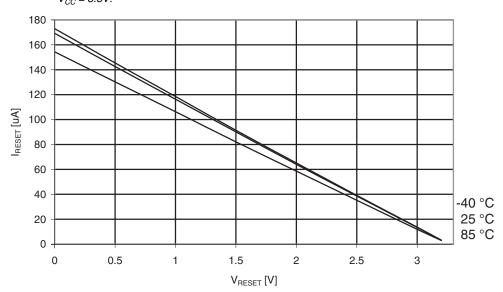
Figure 35-18. Reset Pull-up Resistor Current vs. Reset Pin Voltage





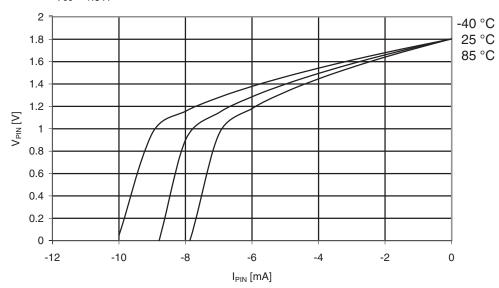
**Figure 35-19.** Reset Pull-up Resistor Current vs. Reset Pin Voltage  $V_{CC} = 3.0V$ .



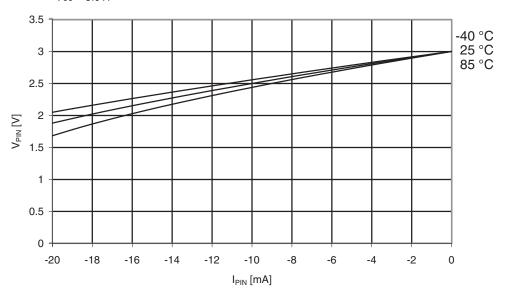


# 35.6 Pin Output Voltage vs. Sink/Source Current

**Figure 35-21.** I/O Pin Output Voltage vs. Source Current Vcc = 1.8V

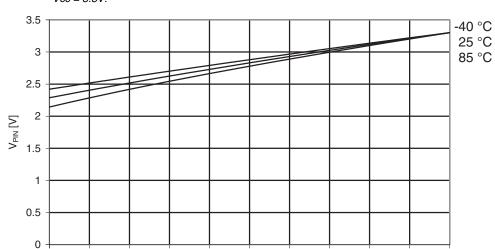


**Figure 35-22.** I/O Pin Output Voltage vs. Source Current Vcc = 3.0V.



0

-2



-10

I<sub>PIN</sub> [mA]

-8

-6

-4

**Figure 35-23.** I/O Pin Output Voltage vs. Source Current Vcc = 3.3V.

**Figure 35-24.** I/O Pin Output Voltage vs. Sink Current Vcc = 1.8V.

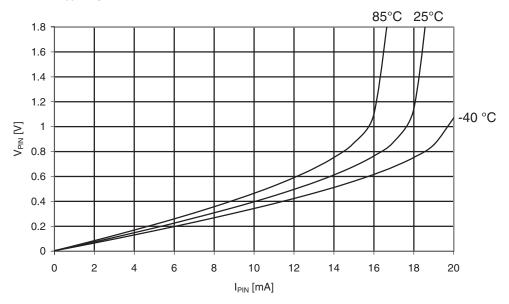
-16

-14

-12

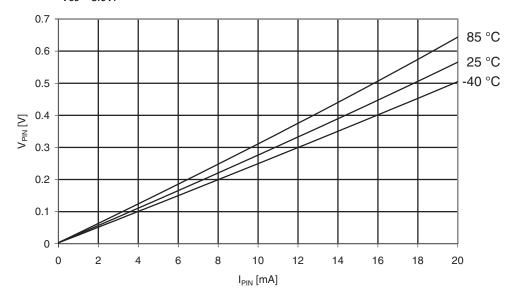
-20

-18

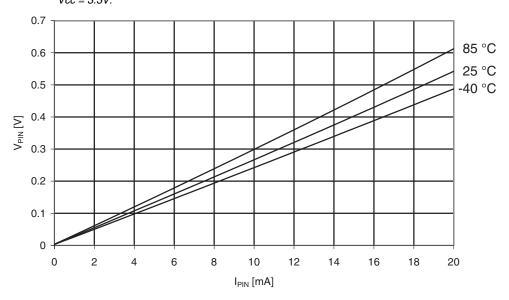




**Figure 35-25.** I/O Pin Output Voltage vs. Sink Current Vcc = 3.0V.



**Figure 35-26.** I/O Pin Output Voltage vs. Sink Current Vcc = 3.3V.



# 35.7 Pin Thresholds and Hysteresis

**Figure 35-27.** I/O Pin Input Threshold Voltage vs.  $V_{CC}$   $V_{IH}$  - I/O Pin Read as "1".

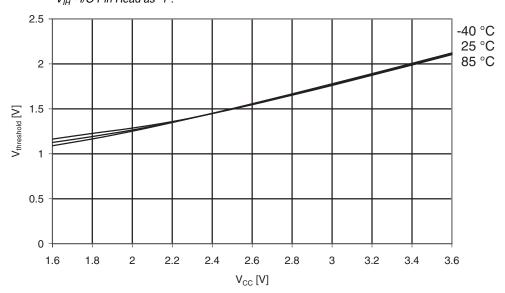
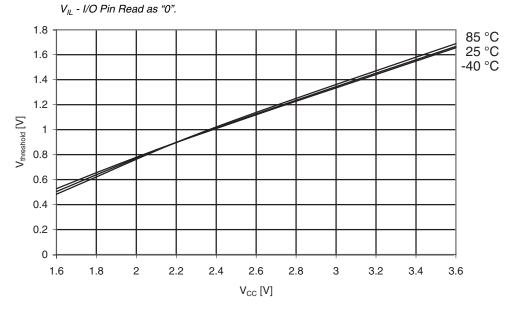


Figure 35-28. I/O Pin Input Threshold Voltage vs.  $\rm V_{\rm CC}$ 





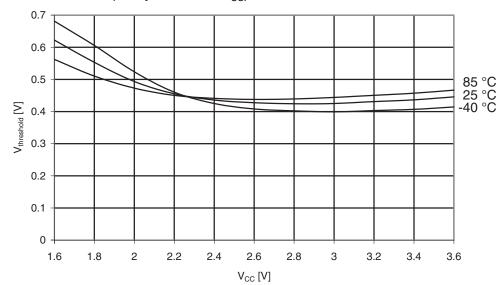
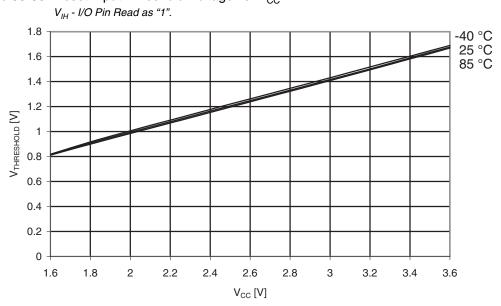


Figure 35-29. I/O Pin Input Hysteresis vs. V<sub>CC.</sub>



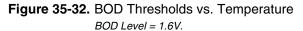


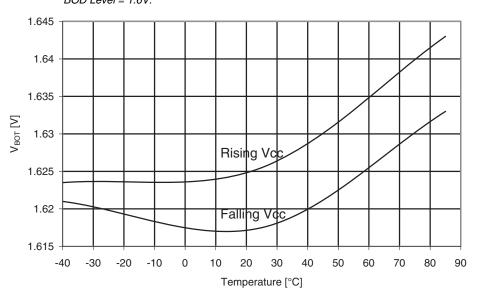


1.8 -40 °C 25 °C 85 °C 1.6 1.4 1.2 V<sub>THRESHOLD</sub> [V] 1 0.8 0.6 0.4 0.2 0 1.6 1.8 2 2.2 2.4 2.6 2.8 3 3.2 3.4 3.6  $V_{CC}[V]$ 

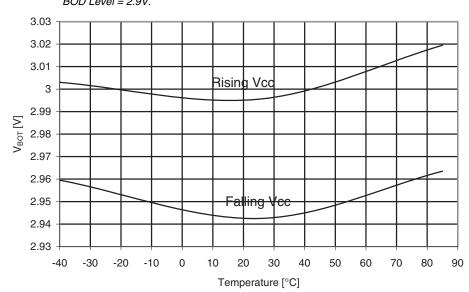
Figure 35-31. Reset Input Threshold Voltage vs.  $V_{CC}$   $V_{IL}$  - I/O Pin Read as "0".

### 35.8 Bod Thresholds





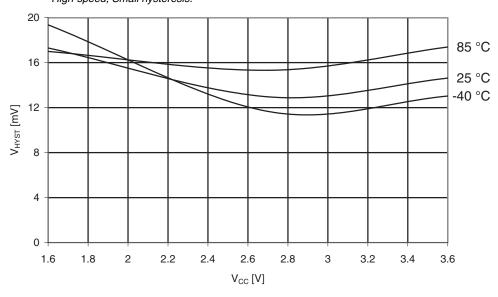




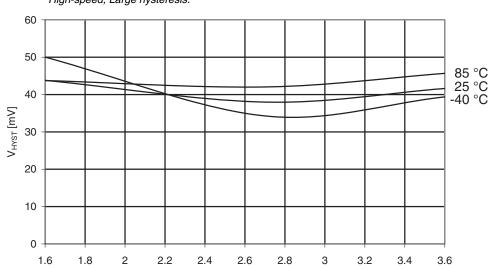
**Figure 35-33.** BOD Thresholds vs. Temperature BOD Level = 2.9V.

# 35.9 Analog Comparator

**Figure 35-34.** Analog Comparator Hysteresis vs. V<sub>CC</sub> *High-speed, Small hysteresis.* 



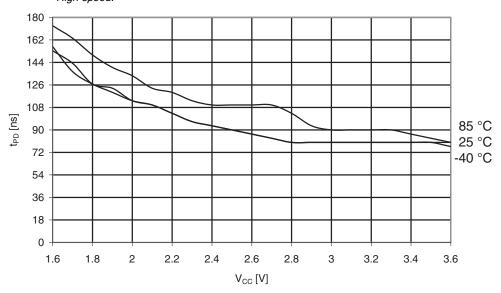




V<sub>CC</sub> [V]

**Figure 35-35.** Analog Comparator Hysteresis vs. V<sub>CC</sub> *High-speed, Large hysteresis.* 

**Figure 35-36.** Analog Comparator Propagation Delay vs. V<sub>CC</sub> *High-speed.* 

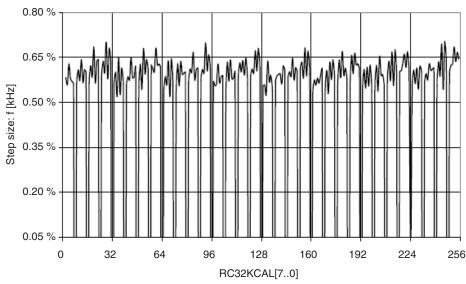




# 35.10 Internal Oscillator Speed

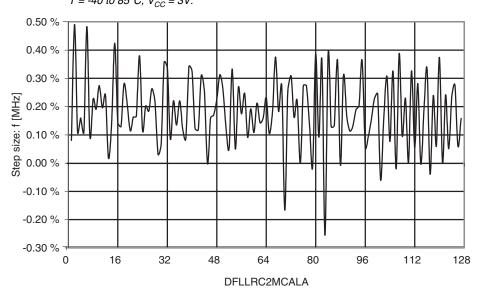
### 35.10.1 Internal 32.768 kHz Oscillator

**Figure 35-37.** Internal 32.768 kHz Oscillator Calibration Step Size T = -40 to 85°C,  $V_{CC} = 3V$ .



### 35.10.2 Internal 2 MHz Oscillator

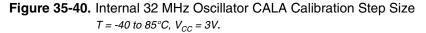
**Figure 35-38.** Internal 2 MHz Oscillator CALA Calibration Step Size T = -40 to 85°C,  $V_{CC} = 3V$ .

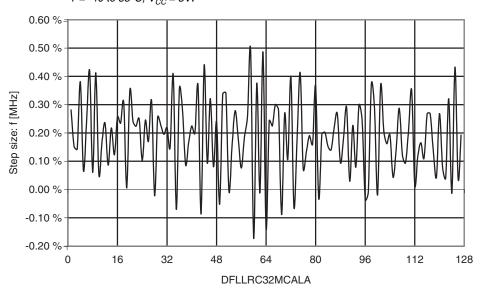


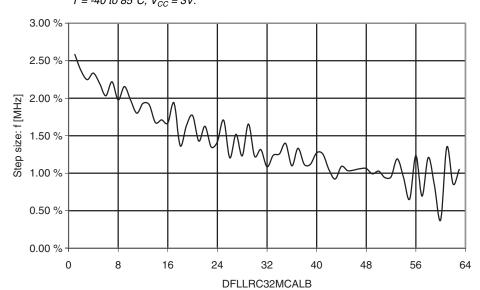
 $T = -40 \text{ to } 85 \,^{\circ}\text{C}, \ V_{CC} = 3V.$ 3.00 % 2.50 % 2.00 % Step size: f [MHz] 1.50 % 1.00 % 0.50 % 0.00 %-0 8 16 24 32 40 48 56 64 DFLLRC2MCALB

Figure 35-39. Internal 2 MHz Oscillator CALB Calibration Step Size

### 35.10.3 Internal 32 MHZ Oscillator



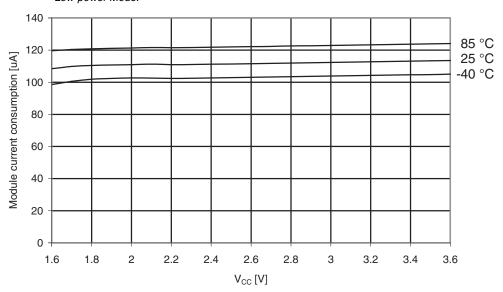




**Figure 35-41.** Internal 32 MHz Oscillator CALB Calibration Step Size T = -40 to  $85^{\circ}C$ ,  $V_{CC} = 3V$ .

# 35.11 Module current consumption

**Figure 35-42.** AC current consumption vs. Vcc *Low-power Mode.* 





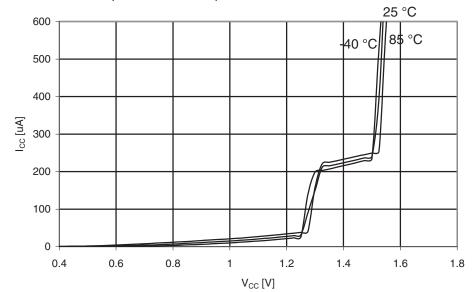


Figure 35-43. Power-up current consumption vs. Vcc

### 35.12 Reset Pulsewidth

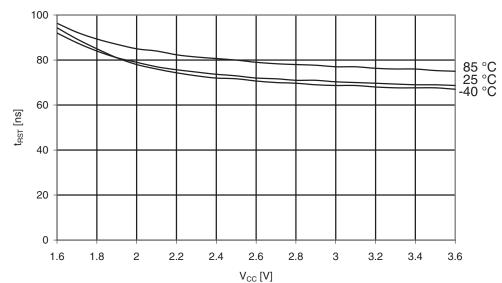


Figure 35-44. Minimum Reset Pulse Width vs. Vcc

### 36. Errata

### 36.1 ATxmega32A4 rev. A

- Flash Power Reduction Mode can not be enabled when entering sleep mode
- Bandgap voltage input for the ACs cannot be changed when used for both ACs simultaneously
- ADC gain stage output range is limited to 2.4V
- Bandgap measurement with the ADC is non-functional when  $V_{\rm CC}$  is below 2.7V
- BOD will be enabled after any reset
- ADC has increased INL error for some operating conditions
- DAC has increased INL or noise for some operating conditions
- · VCC voltage scaler for AC is non-linear

### 1. Flash Power Reduction Mode can not be enabled when entering sleep mode

If Flash Power Reduction Mode is enabled when a deep sleep mode, the device will only wake up on every fourth wake-up request.

If Flash Power Reduction Mode is enabled when entering Idle sleep mode, the wake-up time will vary with up to 16 CPU clock cycles.

#### Problem fix/Workaround

Disable Flash Power Reduction mode before entering sleep mode.

# 2. Bandgap voltage input for the ACs cannot be changed when used for both ACs simultaneously

If the bandgap voltage is selected as input for one Analog Comparator (AC) and then selected/deselected as input for the another AC, the first comparator will be affected for up to 1 us and could potentially give a wrong comparison result.

### Problem fix/Workaround

If the Bandgap is required for both ACs simultaneously, configure the input selection for both ACs before enabling any of them.

#### 3. ADC gain stage output range is limited to 2.4 V

The amplified output of the ADC gain stage will never go above 2.4 V, hence the differential input will only give correct output when below 2.4 V/gain. For the available gain settings, this gives a differential input range of:

```
1x
     gain:
              2.4 V
2x
              1.2 V
     gain:
4x
     gain:
              0.6 V
8x
     gain:
              300 mV
16x
     gain:
              150
                  mV
32x
               75
     gain:
                   mV
64x
               38
                   mV
     gain:
```

#### Problem fix/Workaround

Keep the amplified voltage output from the ADC gain stage below 2.4 V in order to get a correct result, or keep ADC voltage reference below 2.4 V.



### 4. Bandgap measurement with the ADC is non-functional when V<sub>CC</sub> is below 2.7V

The ADC cannot be used to do bandgap measurements when  $V_{CC}$  is below 2.7V.

#### Problem fix/Workaround

If internal voltages must be measured when  $V_{CC}$  is below 2.7V, measure the internal 1.00V reference instead of the bandgap.

### 5. BOD will be enabled after any reset

If any reset source goes active, the BOD will be enabled and keep the device in reset if the VCC voltage is below the programmed BOD level. During Power-On Reset, reset will not be released until VCC is above the programmed BOD level even if the BOD is disabled.

#### Problem fix/Workaround

Do not set the BOD level higher than VCC even if the BOD is not used.

#### 6. ADC has increased INL error for some operating conditions

Some ADC configurations or operating condition will result in increased INL error.

In differential mode INL is increased to:

- 6 LSB for sample rates above 1 Msps, and up to 8 LSB for 2 Msps sample rate.
- 6 LSB for reference voltage below 1.1V when VCC is above 3.0V.
- 20 LSB for ambient temperature below 0 degree C and reference voltage below 1.3V.

In single ended mode, the INL is increased up to a factor of 3 for the conditions above.

#### Problem fix/Workaround

None, avoid using the ADC in the above configurations in order to prevent increased INL error.

### 7. DAC has increased INL or noise for some operating conditions

Some DAC configurations or operating condition will result in increased output error.

- INL error is increased up to 35 LSB when VCC < 2.0V</li>
- Enabling Sample and Hold, will increase noise and reduce resolution below 8 bit

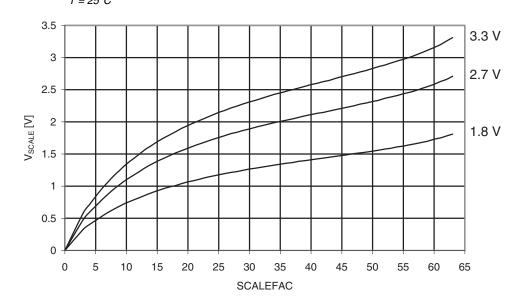
#### Problem fix/Workaround

None, avoid using the DAC in the above configurations in order to prevent increased INL error.

#### 8. VCC voltage scaler for AC is non-linear

The 6-bit VCC voltage scaler in the Analog Comparators is non-linear. The typical voltage output versus the scale factor for different VCC voltages is shown below:





**Figure 36-1.** Analog Comparator Voltage Scaler vs. Scalefac  $T = 25^{\circ}C$ 

#### Problem fix/Workaround

Use external voltage input for the analog comparator if accurate voltage levels are needed

### 36.2 ATxmega16A4 rev. A

- Flash Power Reduction Mode can not be enabled when entering sleep mode
- . Bandgap voltage input for the ACs cannot be changed when used for both ACs simultaneously
- ADC gain stage output range is limited to 2.4V
- $\bullet\,$  Bandgap measurement with the ADC is non-functional when  $V_{CC}$  is below 2.7V
- · BOD will be enabled after any reset
- · ADC has increased INL error for some operating conditions
- DAC has increased INL or noise for some operating conditions
- VCC voltage scaler for AC is non-linear

#### 1. Flash Power Reduction Mode can not be enabled when entering sleep mode

If Flash Power Reduction Mode is enabled when a deep sleep mode, the device will only wake up on every fourth wake-up request.

If Flash Power Reduction Mode is enabled when entering Idle sleep mode, the wake-up time will vary with up to 16 CPU clock cycles.

#### Problem fix/Workaround

Disable Flash Power Reduction mode before entering sleep mode.

# 2. Bandgap voltage input for the ACs cannot be changed when used for both ACs simultaneously

If the bandgap voltage is selected as input for one Analog Comparator (AC) and then selected/deselected as input for the another AC, the first comparator will be affected for up to 1 us and could potentially give a wrong comparison result.



#### Problem fix/Workaround

If the Bandgap is required for both ACs simultaneously, configure the input selection for both ACs before enabling any of them.

### 3. ADC gain stage output range is limited to 2.4 V

The amplified output of the ADC gain stage will never go above 2.4 V, hence the differential input will only give correct output when below 2.4 V/gain. For the available gain settings, this gives a differential input range of:

```
2.4 V
 1x
     gain:
2x
             1.2 V
    gain:
4x
    gain:
              0.6 V
8x
    gain:
             300 mV
16x
    gain:
             150 mV
32x
    gain:
              75
                  mV
64x
              38 mV
    gain:
```

#### Problem fix/Workaround

Keep the amplified voltage output from the ADC gain stage below 2.4 V in order to get a correct result, or keep ADC voltage reference below 2.4 V.

### 4. Bandgap measurement with the ADC is non-functional when V<sub>CC</sub> is below 2.7V

The ADC cannot be used to do bandgap measurements when  $V_{CC}$  is below 2.7V.

#### Problem fix/Workaround

If internal voltages must be measured when  $V_{CC}$  is below 2.7V, measure the internal 1.00V reference instead of the bandgap.

#### 5. BOD will be enabled after any reset

If any reset source goes active, the BOD will be enabled and keep the device in reset if the VCC voltage is below the programmed BOD level. During Power-On Reset, reset will not be released until VCC is above the programmed BOD level even if the BOD is disabled.

### Problem fix/Workaround

Do not set the BOD level higher than VCC even if the BOD is not used.

### 6. ADC has increased INL error for some operating conditions

Some ADC configurations or operating condition will result in increased INL error.

In differential mode INL is increased to:

- 6 LSB for sample rates above 1 Msps, and up to 8 LSB for 2 Msps sample rate.
- 6 LSB for reference voltage below 1.1V when VCC is above 3.0V.
- 20 LSB for ambient temperature below 0 degree C and reference voltage below 1.3V.

In single ended mode, the INL is increased up to a factor of 3 for the conditions above.

### Problem fix/Workaround

None, avoid using the ADC in the above configurations in order to prevent increased INL error.



#### 7. DAC has increased INL or noise for some operating conditions

Some DAC configurations or operating condition will result in increased output error.

- INL error is increased up to 35 LSB when VCC < 2.0V
- Enabling Sample and Hold, will increase noise and reduce resolution below 8 bit

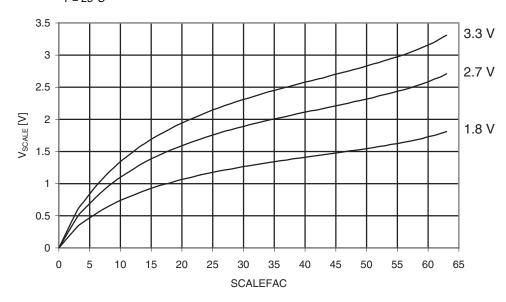
#### Problem fix/Workaround

None, avoid using the DAC in the above configurations in order to prevent increased INL error.

### 8. VCC voltage scaler for AC is non-linear

The 6-bit VCC voltage scaler in the Analog Comparators is non-linear. The typical voltage output versus the scale factor for different VCC voltages is shown below:

**Figure 36-2.** Analog Comparator Voltage Scaler vs. Scalefac  $T = 25^{\circ}C$ 



### Problem fix/Workaround

Use external voltage input for the analog comparator if accurate voltage levels are needed



# 37. Datasheet Revision History

Please note that the referring page numbers in this section are referred to this document. The referring revisions in this section are referring to the document revision.

### 37.1 8069L - 11/09

- 1. Updated "Flash and EEPROM Page Size" on page 14.
- 2. Updated "Electrical Characteristics" on page 61with Max/Min numbers.
- 3. Added "Flash and EEPROM Memory Characteristics" on page 64.
- 4. Updated Table 34-10 on page 67, Input hysteresis is in V and not in mV.

#### 37.2 8069K - 06/09

- 1. Updated "Ordering Information" on page 2.
- 2. Updated "Errata" on page 91

### 37.3 8069J - 04/09

- 1. Updated "Electrical Characteristics" on page 61.
- 2. Updated "Typical Characteristics" on page 69.
- 3. Editorial updates.

### 37.4 8069I - 03/09

- 1. Updated "Electrical Characteristics" on page 61.
- 2. Updated "Typical Characteristics" on page 69.

### 37.5 8069H - 11/08

- 1. Updated "Ordering Information" on page 2.
- 2. Added VFBGA to "Pinout/Block Diagram" on page 3.
- 3. Updated "Block Diagram" on page 6.
- 4. Updated feature list in "Memories" on page 10.
- 5. Added 49-balls VFBGA to "Packaging information" on page 58.



### 37.6 8069G - 10/08

- 1. Updated "Features" on page 1.
- 2. Updated "Ordering Information" on page 2.
- 3. Replaced the package drawing "44M1" on page 59 by a rev H update.

### 37.7 8069F - 09/08

- Updated "Features" on page 1.
- 2. Updated "Ordering Information" on page 2.
- 3. Updated "Features" on page 10 by removing "External Memory...".
- 4. Updated Figure 7-1 on page 11 and Figure 7-2 on page 12.
- 5. Updated Table 7-2 on page 14 and Table 7-3 on page 14.
- 6. Updated ADC "Features" on page 41 and "Overview" on page 41.
- 7. Removed "Interrupt Vector Summary" section from datasheet.

### 37.8 8069E - 08/08

- 1. Changed Figure 2-1's title to "Bock Diagram and TQFP/QFN pinout".
- 2. Updated Table 30-6 on page 52.

#### 37.9 8069D - 08/08

- 1. Updated "Features" on page 1 and "Overview" on page 5.
- 2. Inserted "Interrupt Vector Summary." on page 52.

### 37.10 8069C - 06/08

- 1. Updated Figure 2-1 on page 3 and "Pinout and Pin Functions" on page 49.
- 2. Updated "Overview" on page 5.
- Updated XMEGA A4 Block Diagram, Figure 3-1 on page 6 by removing JTAG from the block diagram.
- 4. Removed the sections related to JTAG: JTAG Reset and JTAG Interface.
- 5. Updated Table 14-1 on page 25.
- 6. Updated all tables in section "Alternate Pin Functions" on page 51.



### 37.11 8069B - 06/08

- 1. Updated "Features" on page 1.
- 2. Updated "Pinout/Block Diagram" on page 3 and "Pinout and Pin Functions" on page 49.
- 3. Updated "Ordering Information" on page 2.
- 4. Updated "Overview" on page 5, included the XMEGA A4 explanation text on page 6.
- 5. Added XMEGA A4 Block Diagram, Figure 3-1 on page 6.
- 6. Updated AVR CPU "Features" on page 8 and Updated Figure 6-1 on page 8.
- 7. Updated Event System block diagram, Figure 9-1 on page 17.
- 8. Updated "PMIC Programmable Multi-level Interrupt Controller" on page 25.
- 9. Updated "AC Analog Comparator" on page 44.
- 10. Updated "I/O configuration" on page 27.
- 11. Inserted a new Figure 16-1 on page 32.
- 12. Updated "Peripheral Module Address Map" on page 53.
- 13. Inserted "Instruction Set Summary" on page 54.
- 14. Added Speed grades in "Speed" on page 63.

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# **Table of Contents**

	Features	7
	Typical Applications	1
1	Ordering Information	2
2	Pinout/Block Diagram	3
3	Overview	5
	3.1Block Diagram	6
4	Resources	7
	4.1Recommended reading	
5	Disclaimer	7
6	AVR CPU	8
	6.1Features	8
	6.2Overview	8
	6.3Register File	9
	6.4ALU - Arithmetic Logic Unit	9
	6.5Program Flow	9
7	Memories	10
	7.1Features	10
	7.20verview	10
	7.3In-System Programmable Flash Program Memory	11
	7.4Data Memory	12
	7.5Production Signature Row	13
	7.6User Signature Row	13
	7.7Flash and EEPROM Page Size	14
8	DMAC - Direct Memory Access Controller	15
	8.1Features	15
	8.2Overview	15
9	Event System	16
	9.1Features	16
	9.2Overview	16
10	System Clock and Clock options	18
	10.1Features	18



	10.20verview	18
	10.3Clock Options	19
11	Power Management and Sleep Modes	
	11.1Features	21
	11.2Overview	21
	11.3Sleep Modes	21
12	System Control and Reset	23
	12.1Features	23
	12.2Resetting the AVR	23
	12.3Reset Sources	23
13	WDT - Watchdog Timer	24
	13.1Features	24
	13.2Overview	24
14	PMIC - Programmable Multi-level Interrupt Controller	2
	14.1Features	25
	14.20verview	25
	14.3Interrupt vectors	25
15	I/O Ports	27
	15.1Features	27
	15.2Overview	27
	15.3I/O configuration	27
	15.4Input sensing	30
	15.5Port Interrupt	30
	15.6Alternate Port Functions	30
16	T/C - 16-bit Timer/Counter	<b>3</b> 1
	16.1Features	31
	16.20verview	31
17	AWEX - Advanced Waveform Extension	<i>3</i> 3
	17.1Features	33
	17.20verview	33
18	Hi-Res - High Resolution Extension	34
	18.1Features	34
	18 20verview	3/



19	RTC - 16-bit Real-Time Counter	<i>35</i>
	19.1Features	35
	19.2Overview	35
20	TWI - Two-Wire Interface	36
	20.1Features	36
	20.2Overview	36
21	SPI - Serial Peripheral Interface	37
	21.1Features	37
	21.2Overview	37
22	USART	38
	22.1Features	38
	22.2Overview	38
23	IRCOM - IR Communication Module	39
	23.1Features	39
	23.2Overview	39
24	Crypto Engine	40
	24.1Features	40
	24.20verview	40
25	ADC - 12-bit Analog to Digital Converter	41
	25.1Features	41
	25.2Overview	41
26	DAC - 12-bit Digital to Analog Converter	43
	26.1Features	43
	26.20verview	43
27	AC - Analog Comparator	44
	27.1Features	44
	27.2Overview	44
	27.3Input Selection	46
	27.4Window Function	46
28	OCD - On-chip Debug	47
	28.1Features	47
	28.20verview	47
29	Program and Debug Interfaces	48



	29.1Features	48
	29.2Overview	48
	29.3PDI - Program and Debug Interface	48
<i>30</i>	Pinout and Pin Functions	49
	30.1Alternate Pin Functions Description	49
	30.2Alternate Pin Functions	51
31	Peripheral Module Address Map	53
32	Instruction Set Summary	54
33	Packaging information	58
	33.144A	58
	33.244M1	59
	33.349C2	60
34	Electrical Characteristics	61
	34.1Absolute Maximum Ratings*	61
	34.2DC Characteristics	61
	34.3Speed	63
	34.4Flash and EEPROM Memory Characteristics	64
	34.5ADC Characteristics	65
	34.6DAC Characteristics	66
	34.7Analog Comparator Characteristics	66
	34.8Bandgap Characteristics	66
	34.9Brownout Detection Characteristics	67
	34.10PAD Characteristics	67
	34.11POR Characteristics	68
	34.12Reset Characteristics	68
	34.13Oscillator Characteristics	68
<i>35</i>	Typical Characteristics	69
	35.1Active Supply Current	69
	35.2Idle Supply Current	72
	35.3Power-down Supply Current	76
	35.4Power-save Supply Current	77
	35.5Pin Pull-up	77
	35.6Pin Output Voltage vs. Sink/Source Current	79
	35.7Pin Thresholds and Hysteresis	82



	35.8Bod Thresholds84
	35.9Analog Comparator85
	35.10Internal Oscillator Speed87
	35.11Module current consumption89
	35.12Reset Pulsewidth90
<i>36</i>	Errata91
	36.1ATxmega32A4 rev. A91
	36.2ATxmega16A4 rev. A93
<i>37</i>	Datasheet Revision History96
	37.18069L – 11/0996
	37.28069K – 06/0996
	37.38069J - 04/0996
	37.480691 – 03/0996
	37.58069H – 11/0896
	37.68069G – 10/0897
	37.78069F – 09/0897
	37.88069E – 08/0897
	37.98069D – 08/0897
	37.108069C - 06/0897
	37.118069B – 06/0898
	37.128069A – 02/0898
	Table of Contentsi



### Headquarters

Atmel Corporation

2325 Orchard Parkway San Jose, CA 95131 USA

Tel: 1(408) 441-0311 Fax: 1(408) 487-2600

#### International

Atmel Asia

Unit 1-5 & 16, 19/F BEA Tower, Millennium City 5 418 Kwun Tong Road Kwun Tong, Kowloon Hong Kong

Tel: (852) 2245-6100 Fax: (852) 2722-1369 Atmel Europe

Le Krebs 8, Rue Jean-Pierre Timbaud BP 309 78054 Saint-Quentin-en-

Yvelines Cedex France

Tel: (33) 1-30-60-70-00 Fax: (33) 1-30-60-71-11

Atmel Japan

9F, Tonetsu Shinkawa Bldg. 1-24-8 Shinkawa Chuo-ku, Tokyo 104-0033

Japan

Tel: (81) 3-3523-3551 Fax: (81) 3-3523-7581

#### **Product Contact**

Web Site

www.atmel.com

**Technical Support** 

avr@atmel.com

Sales Contact

www.atmel.com/contacts

Literature Requests

www.atmel.com/literature

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